

The Initial Parameters Design of the Voltage Source Converter Fed SMES

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ABSTRACT

The initial parameters of the superconducting magnetic energy storage (SMES) fed by a voltage source converter (VSMES) are studied and the setting rules are designed in this paper. A time-domain simulation model is established by using the software PSCAD/EMTDC. Based on this model, the application of the VSMES in the power system is used to test the designed rules. The simulation results are valuable for the further research of the initial parameters design of the VSMES.

Keywords: Superconducting Magnetic Energy Storage (SMES), VSMES, Initial Parameters, Voltage Sag, Time-Domain Simulation

1. Introduction

Superconducting magnetic energy storage system (SMES), a new electric power regulating device used in the power system which combines the superconducting magnet with the power electronic converter, is to store up the electric energy in the form of magnetic energy. If necessary, SMES can compensate the power system for the energy shortage as soon as possible. SMES can not only realize the adjustment of input and output power of SMES in four quadrants, but also control the bi-directional flow of power between the SMES and the power system with a rapid energy exchange rate. Therefore, the SMES is one of important aspects in the application field of superconducting technology [1].

According to the main circuit topology of power electronic converter, the SMES can be classified into the current source and the voltage source. Though current characteristic is the inherent feature of SMES, the voltage source converter, as the mainstream development technology currently, has a lower cost and mature technology. Hence, the SMES system adopts voltage source converter more. Lately, a great deal of literatures had made relative analysis both on the control strategy of SMES and the stability improvement of power system, however, researches concerned on the time-domain modeling and simulation of SMES connected to the power system are rare. This paper establishes the time-domain simulation model of VSMES by the use of PSCAD/ EMTDC, analyses the cases that VSMES compensates the system instantaneous voltage sag and enhances the transient stability of the generator, which have provided references for the practical applications of VSMES in power system [2].

2. Basic Configuration and Work Principle of VSMES

As **Figure 1** shows, the VSMES mainly consists of the voltage source converter, the support capacitance, the chopper, and the superconducting magnet (SM) and is connected to power system through the transformer. The superconducting magnet is equivalent to a pure inductance without resistance and cooperates with the chopper, making the support capacitance a direct current and constant voltage source. VSC is the core of the whole VSMES. Controlling the phase angle and width of trigger pulses of gate turn off thyristor (GTO) in VSC and changing the phase and the amplitude of voltage at the joint of VSMES and the system, which can indirectly control the current between the VSMES and the power system and ultimately realize the control the bidirectional power flow [3].

3. The Time-Domain Simulation Model of VSMES

This paper adopts the user-defined components contained in PSCAD/EMTDC software to build a time-domain digital model that possesses four functions-magnetizing, freewheeling, tracing, demagnetization of VSMES.



Figure 1. Basic construction plan of VSMES

In the model shown as **Figure 2**, the 0.2 Ω resistance and the 0.005 Ω inductance are used to simulate the line loss and the 10 H inductance is used to simulate the superconducting magnet. Es is the phase voltage at the joint of VSMES and the transformer. The voltage source converter employs the three-phase six-impulse bridge inverter. The control pulse of GTO is G1~G6 in turn. According to the control pulse signal produced by the target power, changing the amplitude and the phase of Es can realize the control of the power.

The power control system (PCS) of VSMES adopts double closed-loop serial control, including a voltage outer loop and a current inner loop. Since the direct current control strategy of current inner loop is introduced, the dynamic response of VSMES is improved to a large extent and also the anti-disturbance ability. Through the d-q transform of three-phase voltage and based on the target tracing power, the current value in the d-q coordinate is obtained, moreover, the target current value needed to be regulated can be derived from the decoupling transform of the current above, which achieves the purpose of direct current control [4].

The chopper comprises two-quadrant chopping circuit which is reversible in voltage. It adopts bang-bang control strategy and maintains the voltage of the support capacitor constant. The control of the converter aims at power controlling, while the control of the chopper aims at voltage invariableness which supports the capacitance. In this way, it decreases the difficulty of dynamic matching between the convertor and the chop to a large extent.

4. The Initial Parameters Design of the VSMES

When the disturbance occurs, the VSME can compensate the power system with the storing energy. And the stored energy, in other words the initial parameters of the VSMES decides its compensating capability. So, to design the initial parameters of the VSMES is very important. The initial parameters include L, I_{sm_ref} , U_{dc} and E_s . Where L represents the pure inductance of the superconducting magnet; I_{sm_ref} is the value of initial magnet current; U_{dc} is the real time voltage the supports the capacitance; E_s is the voltage magnitude on AC side.

4.1 The Rule for Selecting L

Superconducting magnet in the superconductive state is equivalent to a pure inductance without resistance in the VSMES. The relationship between the current I, L and energy W is following:

$$W = \frac{1}{2}LI^2 \tag{1}$$

When L is constant, I is in direct proportion to W in (1). But the current carrying capacity of the SM limits the magnitude of I. Meanwhile, Equation (2) shows that the DC voltage on the SM and the SM current all limits the selecting of L.

$$U_{smes} = L \frac{di_{smes}}{dt}$$
(2)

where U_{smes} is the DC voltage on the SM; i_{smes} is the SM current. Considering the above factors and the real SMES made by HUST, the *L* is setting for 10 H in this paper.



Figure 2. The circuit model of VSMES

4.2 The Rule for Selecting *I*_{sm ref}

Considering that the VSMES can compensate the three unbalanced power of the power system during the oscillating period which has the biggest oscillatory power, the reference value $\Delta I'_{sm_ref}$ of the VSMES can be fixed according to the unbalanced power which must be compensated and the compensating time.

$$\Delta E = \frac{1}{2} L \Delta I_{sm_ref}^{\prime 2} \tag{3}$$

where ΔE is the energy which the VSMES stores and just can meet the need of the power compensating.

Supposing the max compensating power by the VSMES during the oscillating period which has the biggest oscillatory power is ΔP_{max} ; and the keeping compensating time is Δt , it must have:

$$\Delta E \approx \Delta P_{\rm max} \times \Delta t \tag{4}$$

It must be pointed out that, when the unbalanced power of the power system likes the low frequency oscillation showed in the **Figure 3**, ΔE must be

$$\Delta E = \frac{\sqrt{2}}{2} \Delta P_{\text{max}} \Delta t \tag{5}$$

Equations (3) and (4) both suppose that the VSMES keeps to supply the energy to the power system during all the compensating time. So, when the fault ends, the SM current is lowest. And the lowest current can be set as I_{sm_min} . Considering the energy loss during the compensating and the loss energy is corresponding to the current ΔI_{sm_loss} varied in the SM, the wanted I_{sm_ref} is:

$$I_{sm_ref} \ge \Delta I'_{sm_ref} + I_{sm_min} + \Delta I_{sm_loss}$$
(6)

In summary, the rule for selecting $I_{sm ref}$ is (7).

$$\begin{cases} \Delta E \approx \Delta P_{\max} \times \Delta t \\ \Delta E = \frac{1}{2} L \Delta I'_{sm_ref} \\ I_{sm_ref} \ge \Delta I'_{ref} + I_{sm_\min} + \Delta I_{loss} \end{cases}$$
(7)



Figure 3. The curve of low frequency oscillation

where, $I_{sm_{min}}$ and $\Delta I_{sm_{loss}}$ need artificially setting and regulating according to the actual simulating result. And $I_{sm_{min}}$ is very critical.

4.3 The Rule for Selecting U_{dc}

In normal condition, the instantaneous adjustable output power by the VSMES meets (8).

$$\sqrt{P_{sm}^2 + Q_{sm}^2} \le U_{dc} I_{smes} \tag{8}$$

where, P_{sm} is the output active power by VSMES; Q_{sm} is the output reactive power by VSMES.

Along with the VSMES's work, its adjustable instantaneous power range gradually is also changing. When the VSMES absorbs energy from the power system, I_{smes} up, $U_{dc}I_{smes}$ up, and the adjustable power range is bigger; otherwise, when the VSMES supplies energy to the power system, I_{smes} down, $U_{dc}I_{smes}$ down, and the adjustable power range is smaller. Meanwhile, the suitable allowance must be considered when selecting $I_{sm_ref.}$ If the allowance is not enough, the VSMES cannot provide the needed power to the power system. In another word, the needed power exceeds the adjustable power range of the VSMES. This will cause the whole VSMES system crash. And I_{sm_min} represents the allowance of I_{sm_ref} actually. So, it is the reason why I_{sm_min} is very critical. It must meet (9).

$$\sqrt{P_{sm}^2 + Q_{sm}^2} \le U_{dc} I_{sm_\min}$$
⁽⁹⁾

According to the analysis above, supposing the system instantaneous power shortage which is needed to compensate is S, the rule for selecting U_{dc} is (10).

$$U_{dc} \approx \frac{S}{I_{sm\ min}} \tag{10}$$

4.4 The Rule for Selecting E_S

In order to ensure the power regulation of the VSMES in the controlled linear range, the signal modulation ratio M in the SPWM modulation must belong to 0 to 1. The relevant calculation of M is (11). And according to the U_{dc} and the P and Q which are the power needed to output by the VSMES, E_s can be got.

Maintaining *M* at 0.5, the corresponding E_s can be got by the method showed in **Figure 4**. It is difficult to deduct E_s directly from U_{dc} according to (11). So the inverse method showed in **Figure 4** is adopted. Presetting the E_s , to calculate U'_{dc} , and then compare it with the setted U_{dc} , judging its suitability by the setting E_s . If U'_{dc} and U_{dc} are more or less the same, then select the value of E_s ; if U'_{dc} and U_{dc} are very different, then continue to adjust the E_s .



Figure 4. The method of selecting E_S

$$\begin{split} M &= 2\sqrt{\left(U_{rd}^{2} + U_{rq}^{2}\right) \times \frac{2}{3}} / U_{dc} \\ \begin{cases} U_{rd} &= U_{rd1} + U_{rd2} + U_{rd3} \\ U_{rq} &= U_{rq1} + U_{rq2} + U_{rq3} \end{cases} \\ \begin{cases} U_{rd1} &= U_{sd} \quad U_{rd2} = \omega Li_{q} \quad U_{rd3} = -Ri_{d} - L\frac{di_{d}}{dt} \\ U_{rq1} &= U_{sq} \quad U_{rq2} = \omega Li_{d} \quad U_{rq3} = -Ri_{q} - L\frac{di_{q}}{dt} \end{cases} \\ \begin{cases} \frac{PU_{sd} + QU_{sq}}{U_{sd}^{2} + U_{sq}^{2}} = I_{d} \\ \frac{PU_{sq} - QU_{sd}}{U_{sd}^{2} + U_{sq}^{2}} = I_{q} \end{cases} \\ \begin{cases} U_{sd} &= \sqrt{\frac{3}{2}}E_{s} \\ U_{sq} &= 0 \\ M \in (0,1) \end{cases} \end{split}$$

4.5 Amendment of I_{sm_ref} , U_{dc} and E_S

The above rules of the VSMES initial parameters are the basis general principles, which are based on the VSMES compensating the power system with equivalent three unbalanced power. But in the real applications of VSMES in power system, due to the existence of outer controller that is designed for achieving the set goals, under the control strategy of the outer controller, the compensating by the VSMES is not equivalent three unbalanced power compensating. Based on these circumstances, it must adjust the values of $I_{sm ref}$ and U_{dc} , ac-

cording to the actual situations.

In the permitted extension of I_{sm_ref} and U_{dc} , it should increase U_{dc} comparatively and reduce I_{sm_ref} . Lower I_{sm_ref} on one hand can help to reduce the internal power loss, on the other hand, help to reduce the design index of current carrying capacity for the SM and the power converter device.

In addition, *Es* is designed according to (11), under the situation that M = 0.5. While in the applications of the VSMES in power system, if Es can be lower and still be able to keep the power conditioning of VSMES in the SPWM linear regulate region, it can set a lower *Es*. This helps reduce the design index of the withstand voltage for the SM and the power converter device.

5. VSMES Used for the Voltage Instantaneous Sag

Figure 5 is the circuit diagram of VSMES compensating the voltage instantaneous sag, in which the rating voltage of the generatrix is 115 kV and the system capability is 500 MVA. When the simulation runs to the 21st s, the system fails for 0.75 s because of the three-phase grounding short circuit. During the 0.75 s, the bus voltage falls from 0.78 (p.u.) to 0.61, then recovers to 0.78.

If the system active power is abundant, VSMES merely need to provide enough reactive power for the system to compensate the voltage instantaneous sag. As **Figure 5** shows, VSMES is connected to the bus through the transformer. The outer loop power control of the VSMES system adopts the bus voltage as the feedback control quantity. Comparing the real-time value of the bus voltage with the rating value, we can get the error signal $\triangle U$ which will be regulated by PI controller to



Figure 5. The circuit diagram of VSMES for compensating the voltage instantaneous sag

generate the reactive power control signal of the VSMES while the active power control signal remains zero [5-7].

According to the aforementioned design rules for the initial parameters of VSMES, the four initial parameters are setting as below:

The *L* is setting for 10 H. The support capacitance voltage of the VSMES is setting for 100 kV and the initial current of the superconducting magnet is setting for 5 kA. The VSMES is connected to the system in parallel through the step-down transformer, and the secondary voltage of the transformer is 6 kV.

Figure 6 is simulation results of the bus voltage. Figure 6(a) depicts real-time comparison curves of the bus voltage during the whole simulation. Figure 6(b) is transient state comparison curves when the VSMES compensates for the voltage sag. Where Vpu_0 represents the system bus voltage without the availability of the VSMES, and Vpu represents the system bus voltage with the VSMES.

It's obvious that the VSMES compensates the instantaneous voltage sag of the system voltage favorably.

Figure 7 presents each monitoring variable of VSMES during the simulation. **Figure 7(a)** provides the real-time power changing curve of VSMES, in which the upper curve represents the active power Psm, and the lower curve represents the reactive power Qsm. **Figure 7(b)** describes the superconducting magnet current. **Figure 7(c)** displays the voltage of the support capacitance.

When the VSMES compensates the system voltage sag, it is found that the system needs about 125 MVar reactive power. Despite the fact that the VSMES is only modulating reactive power, the magnet current will decline due to losses of some components such as the convertor. In **Figure 7(b)**, the magnet current declines from



Figure 6. The generatrix voltage. (a) Comparison graph of the generatrix voltage (p.u.); (b) Transient state comparison graph of the generatrix voltage(p.u.)

the original 5 kA to 1.8 kA during compensating. While **Figure 7(c)** shows that the voltage of the support capacitance maintains invariable after magnetizing.

It's shown that the VSMES can compensate the instantaneous sag of system voltage rapidly and the setting of initial parameters is logical. As shown in this simulation,



Figure 7. The monitoring variables of VSMES. (a) The realtime power of VSMES; (b) The superconducting magnet current; (c) The voltage of the support capacitance

in the end step when the VSMES compensates the voltage decline and outputs 125 MVar reactive power, the magnet current declines to 1.8 kA, however, the mag- net current and the support capacitance voltage of 100 kV can still enable the VSMES adjust the reactive power of 125 MVar, and ensure that the compensating on the voltage sag keeps normal, avoiding the voltage collapse which supports the capacitance. Put it another way, the rules of the initial parameters design of the VSMES is feasible.

However, if the initial parameters are changed not according to the designed rules, for example: the initial current of the superconducting magnet is lowered to 4.5 kA, the support capacitance voltage will decrease when VSMES is compensating the voltage sag as **Figure 8**



Figure 8. The variables of VSMES after initial parameters changed. (a) The superconducting magnet current after initial parameters changed; (b) The voltage of the support capacitance after initial parameters changed

shows.

It's obvious that the energy stored in the superconducting magnet is not enough for the compensating. So it needs additional energy which supported by the support capacitance at the end of the compensating. Then the voltage of the support capacitance decreases. This will have negative effect on the control of VSMES. Therefore, the changed initial parameter is not reasonable. Meanwhile it can be approved that the designed rules in this article are feasible and practical.

6. Conclusions

Based on the principle analysis of the VSMES, this paper establishes a time-domain simulation model of the VSMES with six-pulse and studies on the application of the VSMES in the power system.

To sum up:

1) On the basis of the detailed research on the operational characteristics of the VSMES, the time-domain simulation model of the VSMES is established.

2) The initial parameters setting of the VSMES directly determines the effects on the power system and involves the problem of the power range. This paper proposes a basic rule for parameters setting and uses an example to testify its feasibility.

3) The VSMES can adjust power in four quadrants and

the dynamic response is rapid, which is desirable for compensating the voltage sag of the system.

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