

Antenna Performance Improvement in Elliptical Array Using RMI Method of Mutual Coupling Compensation

Aluru Venkata Lakshmi Narayana Rao¹, Nunna Bala Ankaiah², Dharma Raj Cheruku³

¹Department of ECE, SSNCET, Ongole, India,

²ECIL, Hyderabad, India,

³Departement of ECE, GIT, GITAM University, Visakhapatnam, India

Email: narayanaavlr@gmail.com, bala.nunna8@gmail.com, dharmarajc@yahoo.com

Received 4 December 2015; accepted 25 January 2016; published 29 January 2016

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Abstract

The number of wireless electronic gadgets used in mobile communication, vehicle collision avoidance system, compact radars, etc. is extremely increasing at a rapid rate. Thus, the characteristics of the antennas involved in these gadgets are to be designed very stringently so as to avoid interferences & coupling and to improve compatibility, susceptibility, etc. Compact smart antenna with improved performance is highly essential to meet this challenging scenario. Mutual coupling between various elements of an array is one of the main factors which can be considered for improvement of performance of the antenna. Influence of mutual coupling on performance of the antenna is considered in this paper and various techniques to minimize this effect are presented. Effect of mutual coupling on radiation characteristics of the antenna can be compensated employing various methods like Conventional Mutual Impedance (CMI), Receiving Mutual Impedance (RMI). Analysis is presented as comparison between the two methods for different number of elements in the array. Analysis is also presented for different geometries of the array like circular and elliptical for improved performance. The results show performance improvement in the proposed array for parameters like SNR and Speed of convergence.

Keywords

Antenna Arrays, Elliptical (Oval) Arrays, Mutual Coupling, Compensation Methods, Beam Forming

1. Introduction

Wide spread application of wireless communication technology in daily life demands efficient and reliable sig-

nal transmission. Increased directivity, user capacity, battery life and source separation can be achieved by smart antenna to cope-up the demands. Smart antenna is popularly used in radar systems, wireless communications and in vehicle collision avoidance applications [1]. Antenna array is an integral part of smart antenna system. In smart antenna system, array of elements are designed to achieve required beam form for a given application.

A systematic arrangement of group of radiating elements is called an array. Size of the array and aperture proportionally increases with gain of the antenna [2]. Two methods to increase the aperture size of the array are to increase the number of elements and/or inter element spacing of the array [3]. Computational complexity increases with the number of elements. However, this leads to increase in grating lobes. Power efficiency reduces due to grating lobes [4]. A compromise between these two parameters can be achieved by reorienting the geometry and number of elements in the array.

Linear array is simple in structure and is limited to estimation of the elevation angle (one dimensional) of incoming signal only. In addition, the linear array can only estimate the range of azimuth angle limited by 180° ($-90^\circ < \theta < 90^\circ$) [5]. Hence the accuracy of Direction of Arrival (DOA) estimation is greatly affected when the incoming signal is within the range of end-fire array ($70^\circ < |\theta| < 90^\circ$).

The entire transmitted signal is equal to the sum and reradiated signal from the elements. Similarly, induced current on the antenna element reradiates electromagnetic field which would be received by the other neighboring elements in the array. Such mutual coupling effect is usually considered as a defect which degrades the performance of the array. Two types of coupling compensation methods: Conventional Mutual Impedance (CMI) and Receiving Mutual Impedance (RMI) methods are popular.

2. The Mutual Coupling Analysis

For an antenna array with N elements, receiving an incoming signal, being a plane wave source, the received terminal voltage at the k_{th} antenna element V_k can be expressed as

$$V_k = U_k + W_k \quad (1)$$

U_k = received voltage of k_{th} antenna element by the plane wave source alone

W_k = coupled voltage of the scattered fields from the other antenna elements in the array

The coupled voltage W_k in (1) can be written as [6]-[8]

$$W_k = Z_t^{k1} I_1 + Z_t^{k2} I_2 + \dots + Z_t^{k(k-1)} I_{k-1} + Z_t^{k(k+1)} I_{k+1} \quad (2)$$

where Z_k is the receiving mutual impedance between the k_{th} and the i_{th} antenna elements and I_i is the terminal current at the i_{th} antenna element given by:

$$I_i = V_i / Z_L \quad \text{where } i = 1, 2, \dots, N$$

with Z_L being the terminal load impedance of the antenna elements. Putting (2) and (3) into (1), we have:

$$V_k = U_k + Z_t^{k1} \frac{V_1}{Z_L} + Z_t^{k2} \frac{V_2}{Z_L} + \dots + Z_t^{k(k-1)} \frac{V_{k-1}}{Z_L} + Z_t^{k(k+1)} \frac{V_{k+1}}{Z_L} + \dots + Z_t^{kN} \frac{V_N}{Z_L} \quad (3)$$

The relationship between the uncoupled voltages U_k and the received voltages (*i.e.*, coupled voltages) V_k can be written in a matrix equation as:

$$\begin{bmatrix} U_1 \\ U_2 \\ \vdots \\ U_N \end{bmatrix} = \begin{bmatrix} 1 & \frac{Z_t^{12}}{Z_L} & \dots & \frac{Z_t^{1N}}{Z_L} \\ \frac{Z_t^{21}}{Z_L} & 1 & \dots & \frac{Z_t^{2N}}{Z_L} \\ \vdots & \vdots & \ddots & \vdots \\ \frac{Z_t^{N1}}{Z_L} & \frac{Z_t^{N2}}{Z_L} & \dots & 1 \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ \vdots \\ V_N \end{bmatrix} \quad (4)$$

2.1. Conventional Mutual Impedance Method

A circuit theory approach for reducing or compensating mutual coupling effect is discussed in this work. This

method is called conventional mutual impedance method (CMI) [9]. Common mutual coupling can be easily measured directly or from S-parameters.

The relation between the terminal voltage and current can be given by

$$V_1 = I_1 Z_{K,1} + I_2 Z_{K,2} + \dots + I_i Z_{k,i} + \dots + I_K Z_{K,K} + \dots + I_N Z_{K,N} + V_{OK} \quad (5)$$

Z_L being impedance, relation between voltage and current is given by

$$I_i^t = -\frac{V_i^t}{Z_L} \quad (6)$$

The relationship between the open-circuit voltages and terminal voltages can be written as

$$\begin{bmatrix} 1 + \frac{Z_{11}}{Z_L} & \frac{Z_{12}}{Z_L} & \dots & \frac{Z_{1N}}{Z_L} \\ \frac{Z_{21}}{Z_L} & 1 + \frac{Z_{22}}{Z_L} & \dots & \frac{Z_{2N}}{Z_L} \\ \vdots & \vdots & \ddots & \vdots \\ \frac{Z_{N1}}{Z_L} & \frac{Z_{N2}}{Z_L} & \dots & 1 + \frac{Z_{NN}}{Z_L} \end{bmatrix}_{N \times N} \begin{bmatrix} V_1 \\ V_2 \\ \vdots \\ V_N \end{bmatrix}_{1 \times N} = \begin{bmatrix} V_{01} \\ V_{02} \\ \vdots \\ V_{0N} \end{bmatrix}_{1 \times N} \quad (7)$$

2.2. The Receiving Mutual Impedance

Another circuit theory approach for reducing or compensating mutual coupling effect is receiving mutual impedance method (RMI) [10]. The concept of receiving mutual impedance was suggested [11] as an alternative to the concept of conventional mutual impedance [12]. In fact, the need to handle mutual coupling problems differently in transmitting and receiving arrays has been noted before in mobile communications [13] and [14]. The basic difference between the receiving mutual impedance and the conventional mutual impedance lies in the fundamental causes of the coupling. In the conventional mutual impedance, the coupling is caused by the transmitting current distribution on an antenna while in the receiving mutual impedance; the coupling is caused by the receiving current distribution on an antenna.

The receiving mutual impedance between two antennas is the ratio of the coupled voltage across antenna 1's terminal load Z_{L1} (due to the receiving current distribution on antenna 2) to the terminal current through antenna 2's terminal load Z_{L2} when the array is excited by an external plane wave source.

$$Z_i^{12} = -(V_1 - U_1)/I_2 \quad (8)$$

i.e. Receiving mutual impedance with a receiving current on antenna 2 = coupled voltages across antenna 1's terminal load/receiving current through antenna 2's terminal load.

The definition of the receiving mutual impedance requires specifying a plane wave to excite the two antennas as shown in **Figure 1**. The coming direction of the plane wave is to be the same as the direction of the signal that the array is designed. In **Figure 1**, V_1 and V_2 are the received voltages across the terminal loads of antennas 1 and 2, respectively. The corresponding currents through the two antennas are I_1 and I_2 . In equation (1), U_1 is the received voltage across the terminal load of antenna 1 when antenna 1 is excited by the external plane wave source alone (with antenna 2 removed from the array). U_1 is also called the isolation voltage on antenna 1.

3. Geometry and Array Factor

Two different geometries of the array are considered for comparison in this paper viz., Circular and Elliptical (Oval).

3.1. Circular Array

In circular array it is assumed that N equally spaced isotropic elements are placed on X-Y plane along a circular ring. Let the radius be "a". Circular array can able to scan 360° azimuthally. The geometry of N-element circular

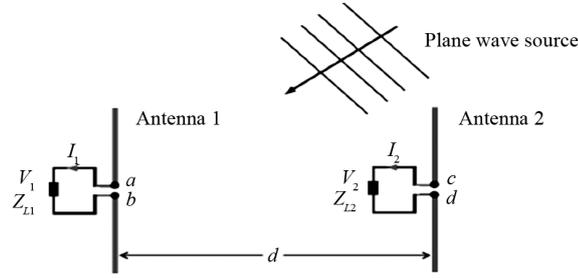


Figure 1. Mutual impedance between antennas.

array antenna is shown in **Figure 2** [15]. The array factor is given by the equation [16]

$$AF(\theta, \phi) = \sum_{N-1}^N I_n e^{j[ka \sin \theta \cos(\phi - \phi_n) + \alpha_n]} \quad (9)$$

where,

I_n = amplitude of excitation

α_n = phase of the n th element

θ = elevation angle from z axis.

Circumference = $2\pi r$

Area = πr^2

The radius of the array increases the directivity of uniform circular array and tends to a value N [17].

3.2. Elliptical Antenna Array

The geometry of the elliptical antenna array with origin as center is shown in **Figure 3** [18]

The array factor is given by equation (10) [19].

$$AF(\theta, \phi) = \sum I_n \exp\left(j\left[k \sin(\theta)\left(a \cos(\theta_n) \cos(\phi) + b \sin(\theta_n) \sin(\phi)\right) + \alpha_n\right]\right) \quad (10)$$

I_n =amplitude of excitation

α_n =phase of the n th element

θ =elevation angle from z axis.

ϕ_n =Azimuth angle measured from x axis for n -th element.

a, b =semi major and minor axes respectively.

e =eccentricity of elliptical array and is 0.5

$$\alpha_n = -k \sin(\theta_0)\left(a \cos(\phi_n) \cos(\phi_0) + b \sin(\phi_n) \sin(\phi_0)\right) \quad (11)$$

$\theta_0 = 90^\circ, \phi_0 = 0^\circ$.

N =no of elements is 8 or 10.

Area (A) = Πab

circumference(c) = $\Pi\left(3(a+b) - \sqrt{(3a+b)(a+3b)}\right)$

Comparisons of area of circular and elliptical arrays are given below in **Table 1**.

4. Results

4.1. Conventional Mutual Impedance Method of Compensation

The objective is to analyze the response of elliptical antenna array with modified LMS (Least Mean Square) algorithm which optimizes the weight factor for DOA (Direction of Arrival) estimation. Mutual coupling is minimized by the CMI method in this case.

Example 1: Eight (Dipole) Element Array using CMI

In this example, an 8-element array using CMI is optimized with uniform spacing between the elements. It is assumed that the angle of signal and interference be 90° and 180° respectively.

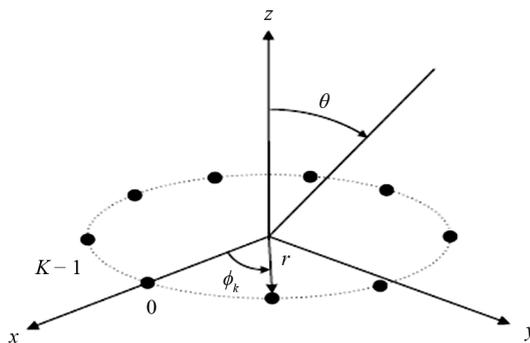


Figure 2. Geometry of circular antenna in XY-plane.

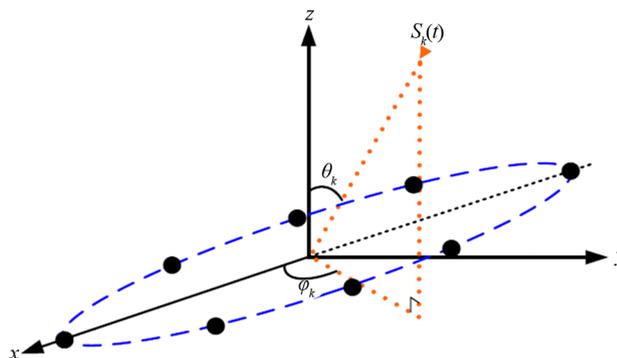


Figure 3. Geometry of ellipse in XY-plane.

Table 1. Comparison of area of circular and elliptical arrays.

Circumference (cm)	Circular and oval array area.				
	Circular array radius r (cm)	Circular array area A (cm)	Oval array major radius a (cm)	Oval array minor radius b (cm)	Area (cm ²)
1.00	0.16	0.08	0.20	0.10	0.07
1.50	0.24	0.18	0.30	0.15	0.15
2.00	0.32	0.32	0.40	0.21	0.27
3.00	0.48	0.72	0.62	0.31	0.60
4.00	0.64	1.27	0.84	0.42	1.07

It can be evident from **Figure 4** and **Figure 5**, that elliptical (oval) array with CMI method has reduced number of side lobes. Also, lower response of -20 dB observed in the interference direction, compared to circular antenna with -10 dB response.

Figure 6 and **Figure 7** compare the output noise in circular and elliptical antennas. Elliptical antenna has noise power which is less by 0.5 dB (negligible noise) for the iterations 10 to 100, whereas the circular antenna has considerable noise at low iteration rates below 20 for minimum of 0.5 dB. Hence it can be concluded that convergence is fast in elliptical array antenna than in circular for low noise.

Example 2: Ten (Dipole) Element Array using CMI

In this example, a 10-element array using CMI is optimized with uniform spacing between the elements. It is assumed that the angle of signal and interference be 90° and 180° respectively.

In **Figure 8**, iterations vs. SNR in circular array with 10-elements (uniform spacing between elements) using RMI Method is presented. **Figure 9** is similar plot for Elliptical array. The elliptical antenna leads with better SNR than circular antenna for iterations below 20. Circular antenna has -10 dB SNR and elliptical antenna has -2 dB which confirms the faster convergence of elliptical array.

Figure 10 shows circular array with CMI Method and without CMI Method which has deeper nulls if compensation is not applied. But elliptical array performance is good with null points covered by peaks as shown in **Figure 11**.

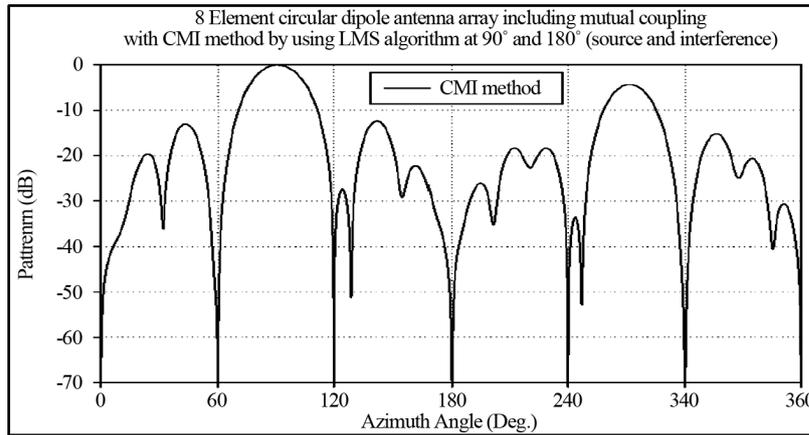


Figure 4. Circular array using CMI Method.

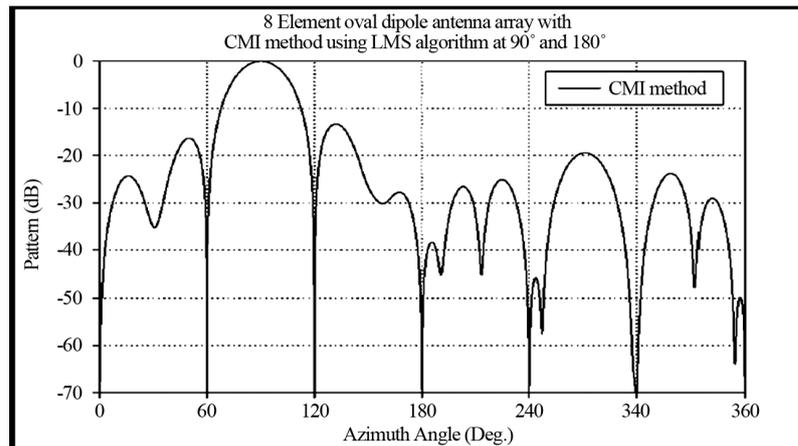


Figure 5. Elliptical (oval) array with CMI Method.

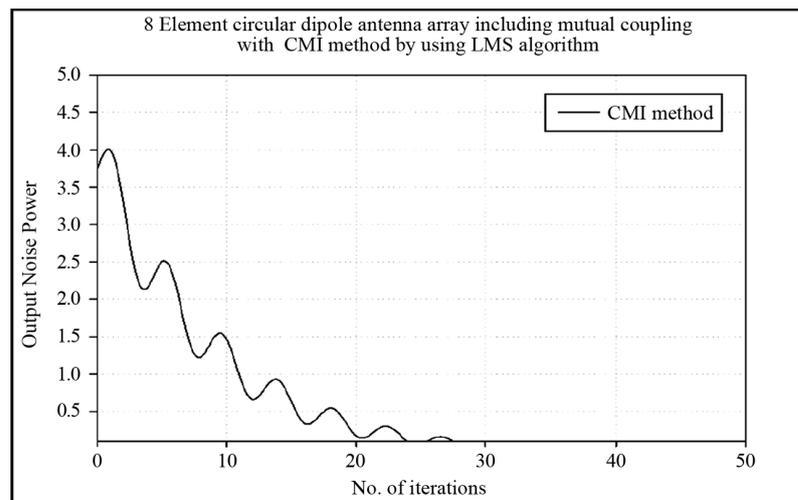


Figure 6. Iterations vs. output noise in circular array with CMIMethod.

In CMI compensation both 8 & 10 element arrays of circular and elliptical arrays are compared. Elliptical arrays show better in pattern generation, SNR and convergence speed. Advantage of elliptical antenna compared to circular antenna is its low area for same number of elements.

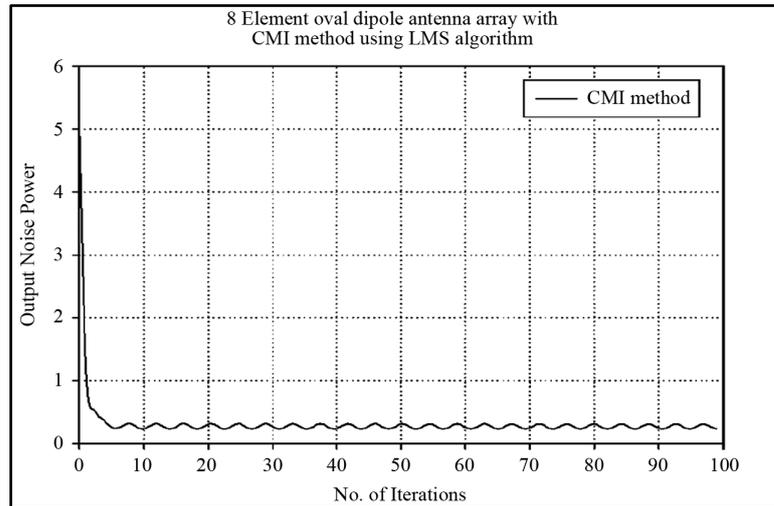


Figure 7. Iterations vs. output noise in Elliptical(oval) array using CMI Method.

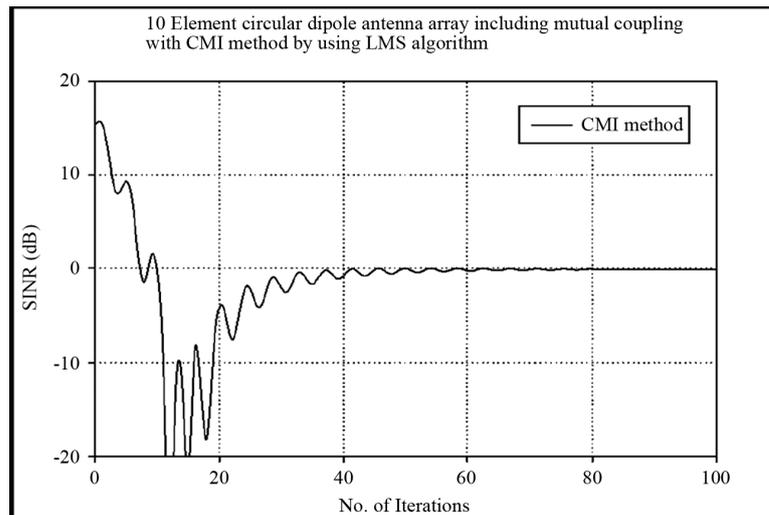


Figure 8. Iterations vs. SINR in circular array with CMI Method.

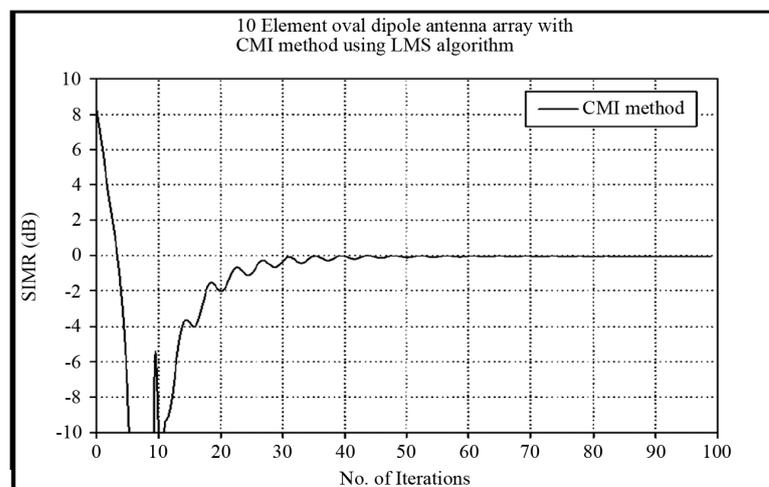


Figure 9. Iterations vs. output SNR in Elliptical array using CMI Method.

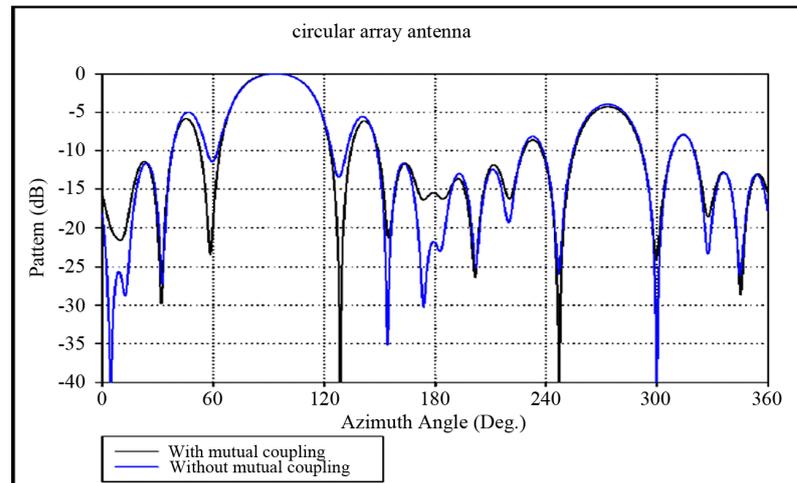


Figure 10. Circular array with and without CMI Method.

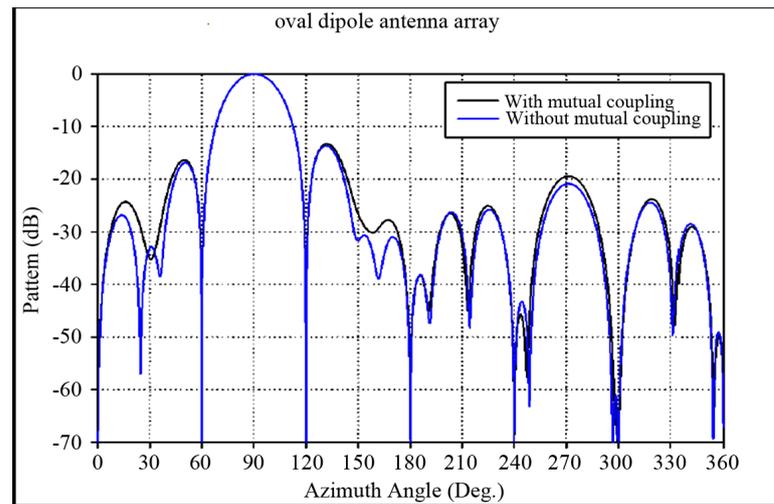


Figure 11. Elliptical array with and without CMI Method.

4.2. The Receiving Mutual Impedance Method of Compensation

In this RMI Method, the source and noise are assumed to be 90° and 180° of azimuth directions respectively. A modified LMS algorithm is used to optimize the weight factor for DOA estimation.

Example 3: Eight (Dipole) Element Array using RMI

In this example, an 8-element Circular array is optimized with uniform spacing between the elements.

The best amplitude value determined by the optimized technique is shown in Figure 12. From Figure 13 it can be noted that the output noise power sustains minimum with the number of iterations above 10.

It can be evident that from Figure 13 elliptical array with RMI Method has reduced side lobes and peaks compared with circular array with RMI Method in Figure 12. Elliptical antenna has lower response in the interference direction than circular antenna.

Figure 14 and Figure 15 compare the output noise in circular and elliptical arrays. Elliptical antenna has negligible noise throughout the iterations from 10 to 100 below 0.5 dB, whereas the circular antenna has noise slightly more than that of elliptical antenna. Also below 10 iterations makes much difference in both antenna output noise. Convergence is faster in elliptical antenna.

Example 4: Ten (Dipole) Element Array using RMI

In this example, a 10-element array with uniform spacing between the elements is optimized using RMI. As per the Figure 16 & Figure 17 for Iterations vs. SNR in circular and elliptical arrays (no of elements 10) using

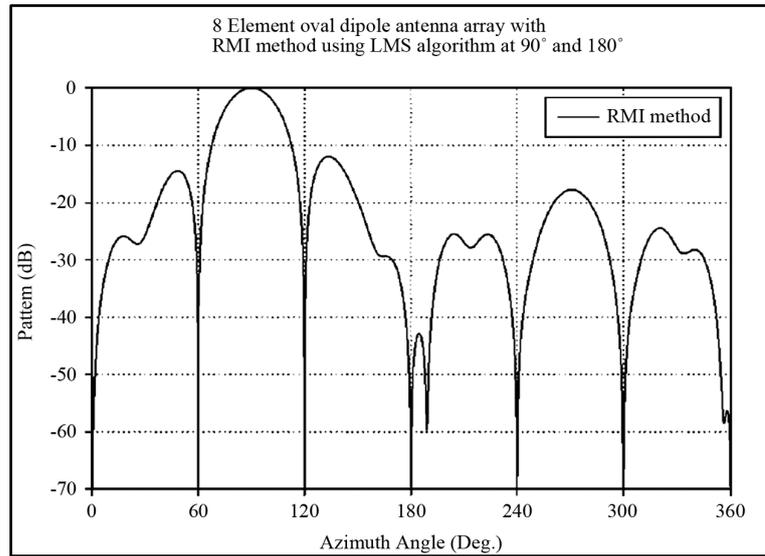


Figure 12. Circular array with RMI Method.

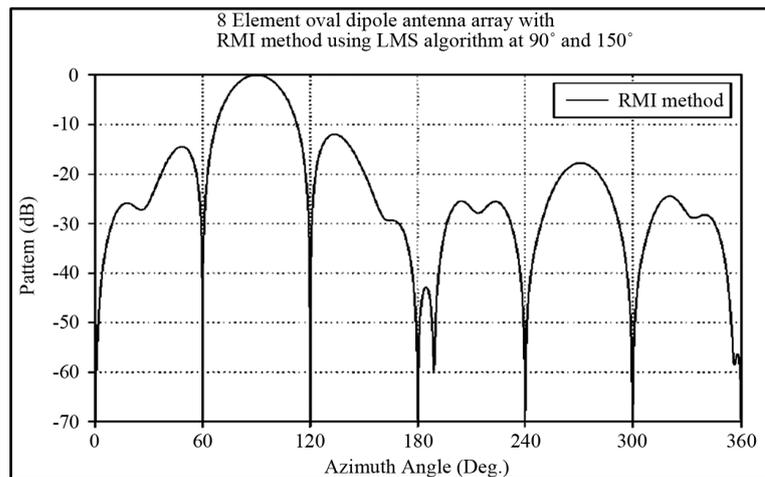


Figure 13. Elliptical (oval) array with RMI Method.

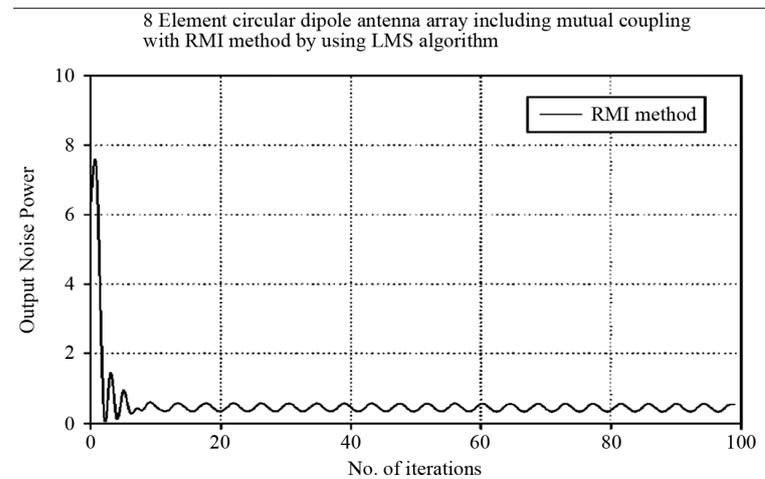


Figure 14. Iterations vs. output noise in circular array with RMI Method.

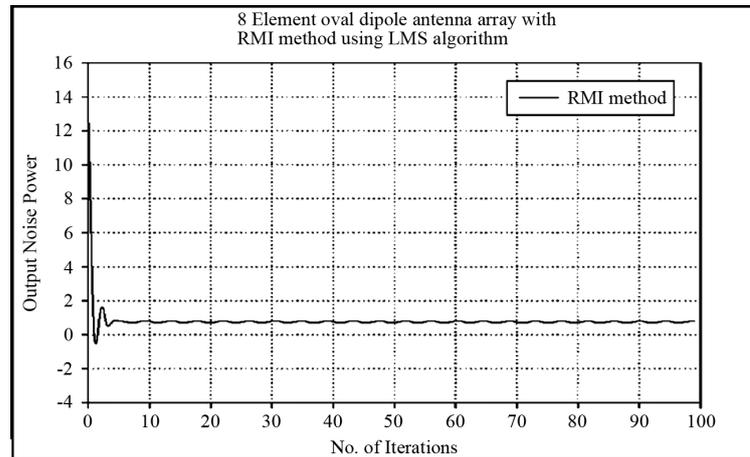


Figure 15. Iterations vs output noise in Elliptical (oval) array using RMI Method.

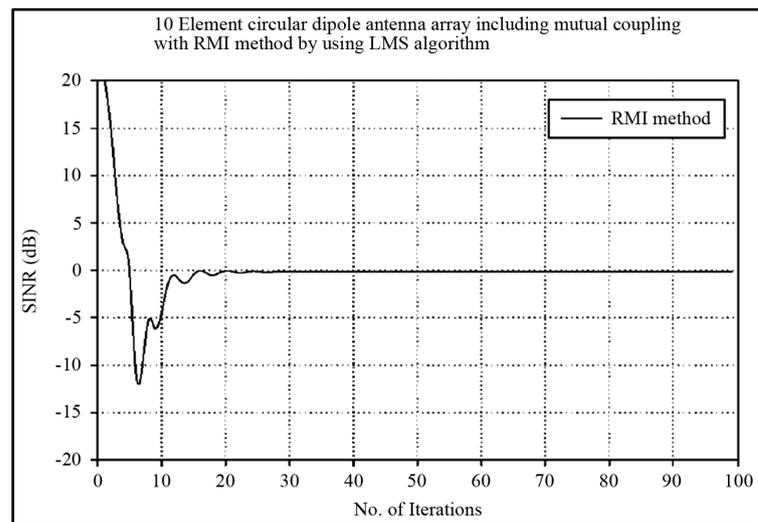


Figure 16. Iterations vs. SINR in circular array with RMI Method.

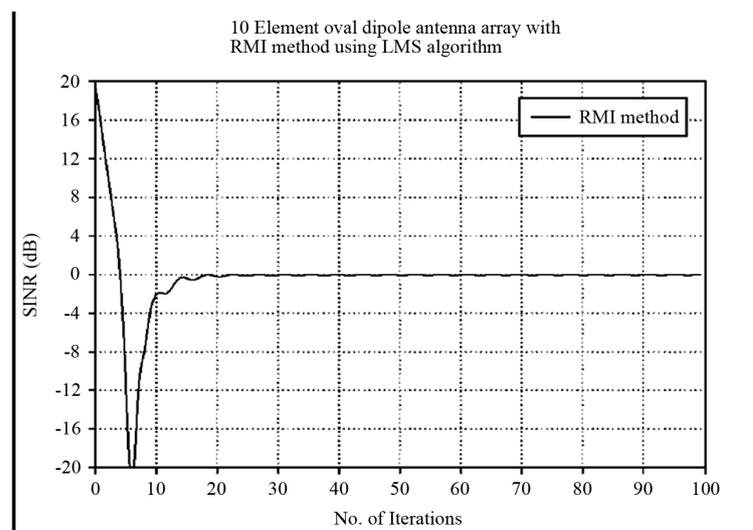


Figure 17. Iterations vs. output SINR in Elliptical array using RMI Method.

RMI Method, the elliptical antenna has better SNR below 20 iterations. Hence convergence is faster in elliptical antenna. For the iterations below 10, circular array has minimum of -12 db SNR.

Figure 18 and **Figure 19** presents the comparison between the patterns generated with RMI and without RMI method for circular array and elliptical antennas. Elliptical antenna beam minimized lobe in 120° to 240° , azimuth angle compared to circular antenna.

Example 5: Sixteen (Dipole) Element Array using RMI

In this example, an 16-element Circular & Elliptical arrays are optimized with uniform spacing between the elements. Simulations are shown in **Figure 20** & **Figure 21**. It observed that from **Figure 21** Elliptical (oval) array with RMI Method has reduced side lobes and peaks compared with circular array. Elliptical antenna has peak at -28 dB in the interference direction as circular antenna has -21 dB

In RMI compensation along with 8 & 10 element arrays of circular and elliptical arrays, 16-element arrays are also compared. Elliptical arrays shows better in pattern generation, SNR. Convergence speed is more improved. Geometrical advantage of compact size for elliptical antenna with same number of elements as in circular antenna is sustained.

More over RMI sanctions receiving current distribution on an antenna compared to CMI mode of compensation. It is additional advantage to recommend the RMI in elliptical arrays so that DOA estimation depends on Direction of receiving signal strength.

Example 6: Sixteen (Dipole) Element Array with non linear spacing using RMI

Figure 22 and **Figure 23** compares the output of 16-element circular & elliptical antenna with non linear spacing. Elliptical antenna has lower response at 210° , 270° and 330° Azimuth direction, whereas the circular antenna has response better than that of elliptical antenna. Also, side lobes are more in circular antenna.

Lobes between $180^\circ - 240^\circ$ & $300^\circ - 330^\circ$ are smoothening and shaped compared to circular array in elliptical antenna.

However, below 180° of angle uniform spaced array is seems to be performing better.

5. Conclusions

The simulations categorically show that CMI compensation method applied to elliptical array with 8 & 10 elements has better response in pattern generation, convergence speed and SNR compared with circular array. Compactness of elliptical antenna compared to circular antenna for the same number of elements is additional geometrical advantage.

Using RMI compensation method, a comparison is made for 8, 10 and 16 element array in circular as well as elliptical array. Elliptical arrays show better performance in pattern generation, SNR. Convergence speed is better compared to CMI method of compensation. Moreover, receiving current distribution on an antenna for RMI

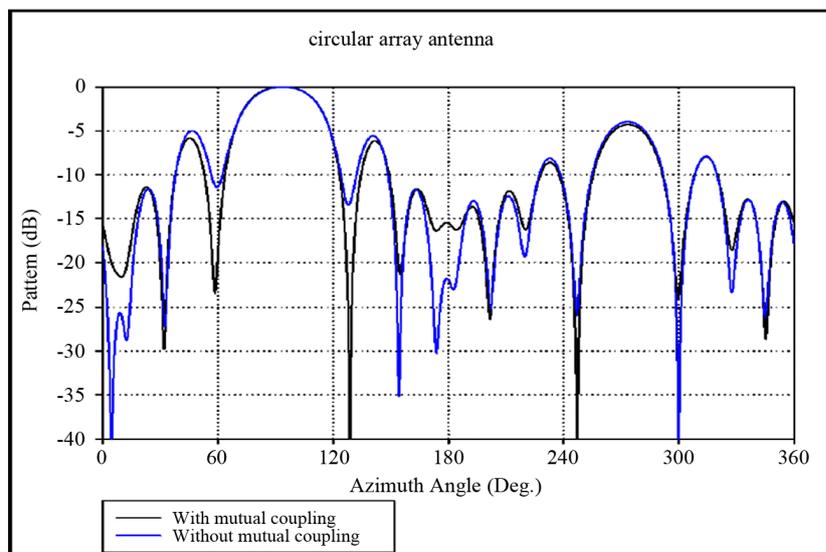


Figure 18. Circular array with and without RMI Method.

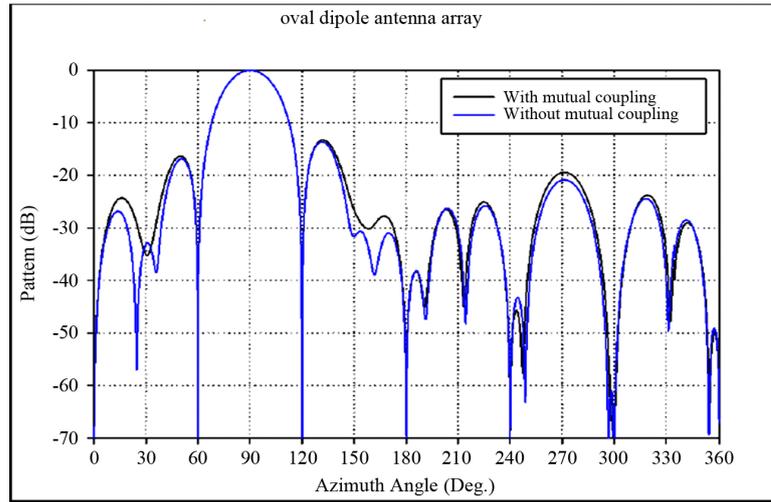


Figure 19. Elliptical array with and without RMI Method.

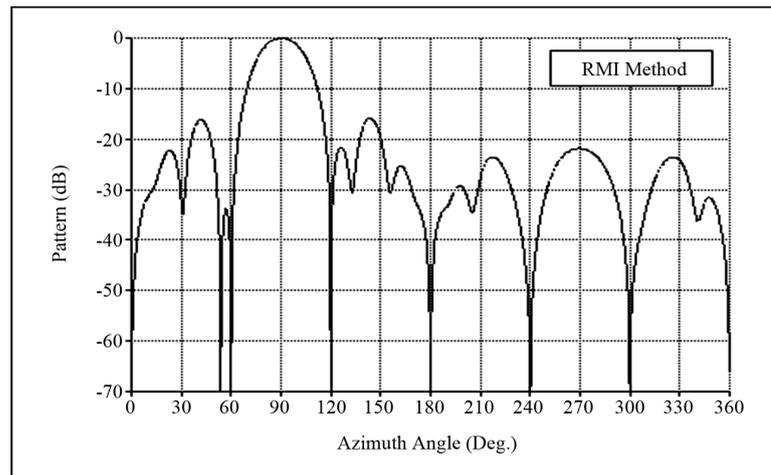


Figure 20. 16-Element Circular Dipole Antenna array with signal and interference at 90° and 180° .

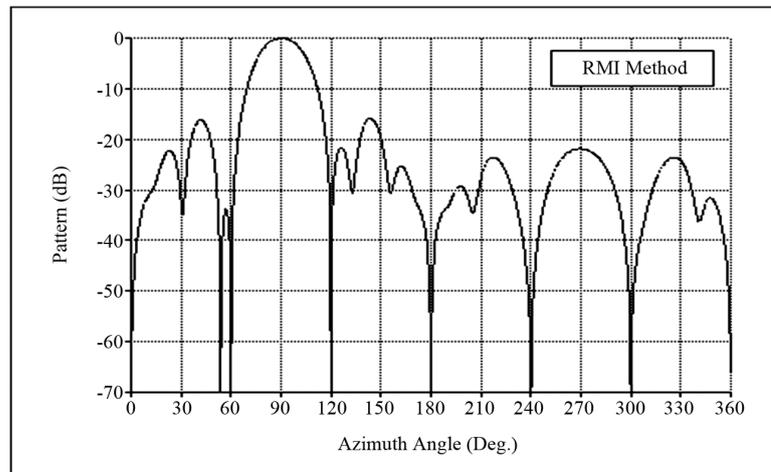


Figure 21. 16-Element elliptical (oval) Dipole Antenna array with signal and interference at 90° and 180° .

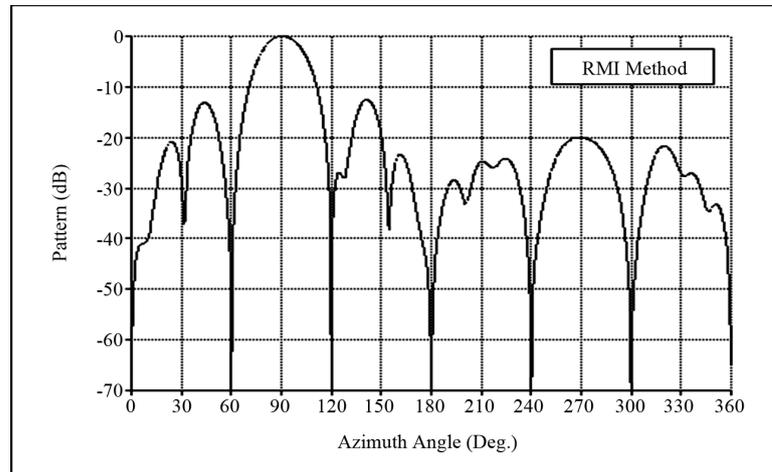


Figure 22. 16-Element Circular Dipole Antenna array with non-uniform signal and interference at 90° and 180° .

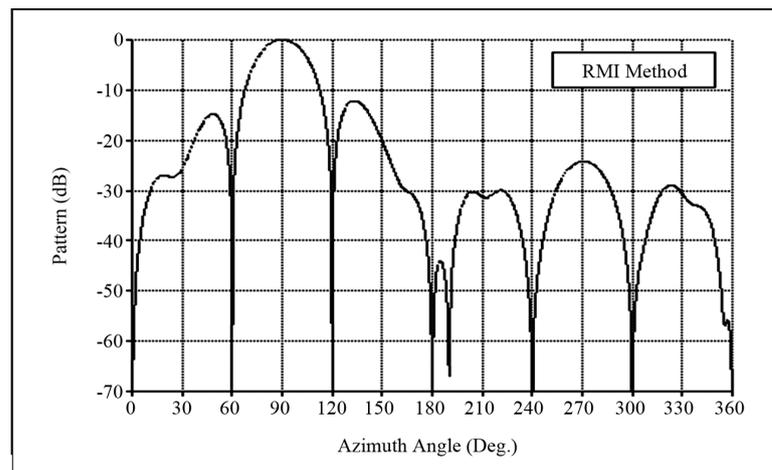


Figure 23. 16-Element Elliptical (oval) Dipole Antenna array with non-uniform signal and interference at 90° and 180° .

method is better compared to CMI mode of compensation. It is additional advantage to recommend the RMI in elliptical arrays because DOA estimation depends on Direction of receiving signal strength.

Also, the simulation of non uniform spaced elements array in either circular or elliptical patterns confirms that below 180 degrees of angle, uniform spaced array performance is better. Hence, it can be concluded that elliptical array of uniform spaced elements with RMI compensation method of mutual coupling has the best performance in a smart antenna.

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