

# Solution of Multi-Delay Dynamic Systems by Using Hybrid Functions

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## Abstract

**In this paper, we present a method for finding the solution of the linear multi-delay systems (MDS) by using the hybrid of the Block-Pulse functions and the Bernoulli polynomials. In this approach, the MDS is reduced to a system of linear algebraic equations by expanding various time functions for the hybrid functions and using operational matrices. To demonstrate the validity and the applicability of the technique, some examples are presented.**

## Keywords

**Block-Pulse Functions, Bernoulli Polynomials, Multi-Delay Systems**

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## 1. Introduction

Delays occur frequently in biological, chemical, transportation, electronic, communication, manufacturing and power systems [1]. Time-delay and multi-delay systems are therefore very important classes of systems whose control and optimization have been of interest to many investigators [2]-[5]. While modeling such phenomena naturally requires the use of various systems, in many problems, such systems can not be solved explicitly. Therefore, it is important to find their approximate solutions by using some numerical methods. In recent years, the hybrid functions consisting of the combination of the Block-Pulse functions with the Chebyshev polynomials [6], the Legendre polynomials [7] [8], or the Taylor series [9] [10] have been shown to be a mathematical power tool for discretization of selected problems. Among these three hybrid functions, hybrids of the Block-Pulse functions with the Legendre polynomials have been shown to be computationally more effective.

Recently a new hybrid function consisting of the combination of the Block-Pulse functions with the Bernoulli polynomials is presented [11] [12]. The advantages of the Bernoulli polynomials  $\beta_m(t)$ ,  $m = 0, 1, 2, \dots$ , over shifted the Legendre polynomials are:

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- The operational matrix  $P$ , in the Bernoulli polynomials, has less errors than  $P$  for shifted the Legendre polynomials for  $1 < m < 10$ . This is because for  $P$  in  $\beta_m(t)$  we ignore the term  $\frac{\beta_{m+1}}{m+1}$  while for  $P$  in  $L_m(t)$  we ignore the term  $\frac{L_{m+1}(t)}{2(2m+1)}$ .
- The Bernoulli polynomials have less terms than shifted the Legendre polynomials. For example  $\beta_6(t)$ , has 5 terms while  $L_6(t)$ , has 7 terms, and this difference will increase by increasing  $m$ . Hence for approximating an arbitrary function we use less CPU time by applying the Bernoulli polynomials as compared to the shifted Legendre polynomials.
- The coefficient of individual terms in the Bernoulli polynomials  $\beta_m(t)$ , is smaller than the coefficient of individual terms in the shifted Legendre polynomials  $L_m(t)$ . Since the computational errors in the product are related to the coefficients of individual terms, the computational errors are less by using the Bernoulli polynomials.

In the present paper, we use the hybrid functions consisting of the combination of the Block-Pulse functions and the Bernoulli polynomials to solve the MDS. The method is based on converting the MDS into a system of multi-delay integral equations through integration. To eliminate integral operations, the unknown functions and various functions involved in the equations are approximated by the hybrid function and the operational matrices are used. To this end, operational matrices of multi-delay systems for the hybrid function are given. It will be seen that the operational matrices have many zero elements and are more sparse than the Legendre polynomials. These matrices are used to reduce the solution of MDS to the solution of a system of linear algebraic equations.

The paper is organized as follows: In Section 2, we describe the basic properties of the hybrid functions of the Block-Pulse and the Bernoulli polynomials required for our subsequent development. Section 3 is devoted to the formulation of linear time-varying multi-delay systems and the proposed numerical method is applied to the MDS. And in Section 4, we report our numerical findings and demonstrate the accuracy of the proposed scheme by considering some numerical examples. Finally, Section 5 gives some brief conclusions.

## 2. Hybrid of the Block-Pulse Functions and the Bernoulli Polynomials

Hybrid functions  $b_{nm}(t), n = 1, 2, \dots, N, m = 0, 1, 2, \dots, M$ , are defined on the interval  $[0, t_f]$  as [11]

$$b_{nm}(t) = \begin{cases} \beta_m(Nt - (n-1)t_f), & t \in \left[ \frac{n-1}{N}t_f, \frac{n}{N}t_f \right], \\ 0 & \text{otherwise,} \end{cases} \quad (1)$$

where  $n$  and  $m$  are the order of the Block-Pulse functions and the Bernoulli polynomials, respectively. The Bernoulli polynomials of order  $m$  are defined in [13] by

$$\beta_m(t) = \sum_{k=0}^m \binom{m}{k} \alpha_k t^{m-k},$$

where  $\alpha_k, k = 0, 1, \dots, m$ , are the Bernoulli numbers. These numbers are a sequence of signed rational numbers that arise in the series expansion of trigonometric functions [14] and can be defined by the identity

$$\frac{t}{e^t - 1} = \sum_{n=0}^{\infty} \alpha_n \frac{t^n}{n!}$$

Let  $f(t)$  be an arbitrary element in  $L^2[0, t_f]$ , there exist unique coefficients  $c_{10}, c_{20}, \dots, c_{NM}$  such that [11]

$$f(t) = \sum_{m=0}^M \sum_{n=1}^N c_{nm} b_{nm}(t) = C^T \phi(t),$$

where

$$C^T = [c_{10}, c_{20}, \dots, c_{N0}, c_{11}, c_{21}, \dots, c_{N1}, \dots, c_{1M}, c_{2M}, \dots, c_{NM}],$$

and

$$\phi(t) = [b_{10}(t), b_{20}(t), \dots, b_{N0}(t), b_{11}(t), b_{21}(t), \dots, b_{N1}(t), \dots, b_{1M}(t), b_{2M}(t), \dots, b_{NM}(t)]^T.$$

By using Equation (2) we obtain

$$f_{ij} = \left\langle \sum_{m=0}^M \sum_{n=1}^N c_{nm} b_{nm}(t), b_{ij}(t) \right\rangle = \sum_{m=0}^M \sum_{n=1}^N c_{nm} k_{nm}^{ij},$$

$i = 1, 2, \dots, N, j = 0, 1, \dots, M,$

where  $f_{ij} = \langle f, b_{ij}(t) \rangle$ ,  $k_{nm}^{ij} = \langle b_{nm}(t), b_{ij}(t) \rangle$ , and  $\langle \cdot, \cdot \rangle$  denotes the inner product. So we get

$$\varphi = K^T C,$$

with

$$\varphi = [f_{10}, f_{20}, \dots, f_{N0}, f_{11}, f_{21}, \dots, f_{N1}, \dots, f_{1M}, f_{2M}, \dots, f_{NM}]^T,$$

$$K = [k_{nm}^{ij}],$$

where  $K$  is a matrix of order  $N(M+1) \times N(M+1)$  and is given by

$$K = \int_0^1 \phi(t) \phi^T(t) dt. \tag{2}$$

Integration of the vector  $\phi(t)$  defined in Equation (4) can be approximated by

$$\int_0^t \phi(t') dt' \approx P \phi(t), \tag{3}$$

where  $P$  is the  $N(M+1) \times N(M+1)$  operational matrix for integration and is given by [11]

$$P = \frac{t_f}{N} \begin{bmatrix} P_0 & I & 0 & \dots & 0 \\ -\frac{1}{2} \alpha_2 I & 0 & \frac{1}{2} I & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ -\frac{1}{M} \alpha_M I & 0 & 0 & \vdots & \frac{1}{M} \\ -\frac{1}{M+1} \alpha_{M+1} I & 0 & 0 & \dots & 0 \end{bmatrix}, \tag{4}$$

where  $I$  and  $0$  are the  $N \times N$  identity and zero matrices, respectively, and

$$P_0 = \begin{bmatrix} -\alpha_1 & 1 & \dots & 1 & 1 \\ 0 & -\alpha_1 & \dots & 1 & 1 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \dots & -\alpha_1 & 1 \\ 0 & 0 & \dots & 0 & -\alpha_1 \end{bmatrix}. \tag{5}$$

The following property of the product of two hybrid function vectors will also be used. Let

$$\phi(t) \phi(t)^T C \approx \tilde{C} \phi(t), \tag{6}$$

where  $\tilde{C}$  is a  $N(M+1) \times N(M+1)$  product operational matrix. To illustrate the calculation procedure see [11].

### Multi-Delay Operational Matrix

The delay functions  $\phi(t - k_j)$ ,  $j = 1, 2, \dots, r$  are the shift of the function  $\phi(t)$  defined in Equation (4), along the time axis by  $k_j$ , where  $k_1, k_2, \dots, k_r$  are rational numbers in  $(0, 1)$ . It is assumed without loss of generality that  $k_1 < k_2 < \dots < k_r$ . If we expand  $\phi(t - k_j)$  in terms of  $\phi(t)$ , we find

$$\phi(t - k_j) = D_j \phi(t), \quad t > k_j,$$

where  $D_j$  is the  $N(M+1) \times N(M+1)$  delay operational matrix of hybrid functions corresponding to  $k_j$  and is given by

$$D_j = \text{diag}(\psi_j, \psi_j, \dots, \psi_j), \quad (7)$$

where elements of the delay matrix are the  $N \times N$  matrix  $\psi_j$  given by

$$\psi_j = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \cdots & 1 \\ 0 & 0 & 0 & \cdots & 0 \end{bmatrix}. \quad (8)$$

It is noted that the first 1 in the first row is located at the  $(\gamma_j + 1)$  th column where

$$\gamma_j = wk_j / \lambda.$$

We define  $w$  as the smallest positive integer number for which  $wk_j \in Z$  for  $j = 1, 2, \dots, r$  and  $\lambda$  is the greatest common divisor of the integers  $wk_j$ ,  $j = 1, 2, \dots, r$ .

### 3. Problem Statement and Approximation Using Hybrid Functions

Consider the following linear time multi-delay dynamic systems:

$$\dot{X}(t) = \sum_{j=1}^r F_j(t) X(t - k_j) + G(t)U(t), \quad 0 \leq t \leq 1, \quad (9)$$

$$X(0) = X_0, \quad (10)$$

$$X(t) = \psi(t), \quad t < 0, \quad (11)$$

where  $X(t) \in R^l$ ,  $U(t) \in R^q$ ,  $G(t)$  and  $F_j(t)$ ,  $j = 1, 2, \dots, r$ , are matrices of appropriate dimensions,  $X_0$  is a constant specified vector, and  $\psi(t)$  is an arbitrary known function. The problem is to find  $X(t)$ ,  $0 \leq t \leq 1$ , satisfying Equations (13) and (14). Let

$$X(t) = [x_1(t), x_2(t), \dots, x_l(t)]^T, \quad U(t) = [u_1(t), u_2(t), \dots, u_q(t)]^T, \quad (12)$$

$$\hat{\phi}(t) = I_l \otimes \phi(t), \quad \hat{\phi}_1(t) = I_q \otimes \phi(t), \quad (13)$$

where  $I_l$  and  $I_q$  are the  $l$ - and  $q$ -dimensional identity matrices,  $\phi(t)$  is  $(M+1)N \times 1$  vector and  $\otimes$  denotes the Kronecker product [15]. Using the property of the Kronecker product,  $\hat{\phi}(t)$  and  $\hat{\phi}_1(t)$  are matrices of order  $l(M+1)N \times l$  and  $q(M+1)N \times q$ , respectively. Assume that each  $x_i(t)$  and each of  $u_j(t)$ ,  $i = 1, 2, \dots, l$ ,  $j = 1, 2, \dots, q$ , can be written in terms of hybrid functions as

$$x_i(t) = \phi^T(t) X_i, \quad u_j(t) = \phi^T(t) U_j.$$

Then, using Equations (15) and (16), we have

$$X(t) = \hat{\phi}^T(t) X, \quad U(t) = \hat{\phi}_1^T(t) U, \quad (14)$$

where  $X$  and  $U$  are vectors of order  $l(M+1)N \times 1$  and  $q(M+1)N \times 1$ , respectively, given by

$$X = [X_1, X_2, \dots, X_l]^T, \quad U = [U_1, U_2, \dots, U_q]^T.$$

Similarly, we have

$$X(0) = \hat{\phi}^T(t) d, \quad \psi(t - k_j) = \hat{\phi}^T(t) R_j, \quad (15)$$

where  $d$  and  $R_j$ ,  $j = 1, 2, \dots, r$ , are vectors of order  $l(M+1)N \times 1$  given by

$$d = [d_1, d_2, \dots, d_l]^T, \quad R_j = [\alpha_{j1}, \alpha_{j2}, \dots, \alpha_{jl}]^T.$$

Let approximate  $G(t)$  and  $F_j(t)$ ,  $j = 1, 2, \dots, r$ , by Equations (2)-(4) as follows

$$G(t) = \phi^T(t)G, \quad F_j(t) = \phi^T(t)F_j, \tag{16}$$

where  $F_j$ ,  $j = 1, 2, \dots, r$  are of dimensions  $l \times l(M+1)N$ , and  $G$  is of dimension  $l \times q(M+1)N$ .

We can also write  $X(t - k_j)$ ,  $j = 1, 2, \dots, r$ , in terms of the hybrid functions as

$$X(t - k_j) = \begin{cases} \hat{\phi}^T(t)R_j & 0 \leq t \leq k_j, \\ \hat{\phi}^T(t)\hat{D}_j^T X & k_j \leq t \leq 1, \end{cases}$$

where

$$\hat{D}_j = I_l \otimes D_j,$$

and  $D_j$  is the delay operational matrix. Moreover

$$\int_0^t \hat{\phi}^T(s) ds = (I_l \otimes \phi^T(t))(I_l \otimes P^T) = \hat{\phi}^T(t)\hat{P}^T, \tag{17}$$

$$\int_0^t F_j(s)X(s - k_j) ds = \begin{cases} \hat{\phi}^T(t)\hat{P}^T\tilde{F}_j^T R_j & 0 \leq t \leq k_j, \\ \hat{\phi}^T(t)Z_j\tilde{F}_j^T R_j + \hat{\phi}^T(t)\hat{P}^T\tilde{F}_j^T\hat{D}_j^T X & k_j \leq t \leq 1, \end{cases} \tag{18}$$

where

$$\hat{P} = I_l \otimes P,$$

and

$$\int_0^{k_j} \hat{\phi}^T(t) dt = \hat{\phi}^T(t)Z_j,$$

where  $Z_j$ ,  $j = 1, 2, \dots, r$ , are constant matrices of order  $l(M+1)N \times l(M+1)N$ . Note  $\tilde{F}_j$  is  $N(M+1) \times N(M+1)$  product operational matrix that to illustrate the calculation procedure we choose  $M = 7$  and  $N = 3$ . Thus we have

$$F_j = [f_{10}^j, f_{20}^j, f_{30}^j, f_{11}^j, \dots, f_{36}^j, f_{17}^j, f_{27}^j, f_{37}^j]^T,$$

$$\tilde{F}_j = \begin{bmatrix} \tilde{F}_{j_0} & \tilde{F}_{j_1} & \tilde{F}_{j_2} & 0 & 0 & 0 & 0 & 0 \\ \frac{1}{12}\tilde{F}_{j_1} & \tilde{F}_{j_0} + \frac{1}{6}\tilde{F}_{j_2} & \tilde{F}_{j_1} & \tilde{F}_{j_2} & 0 & 0 & 0 & 0 \\ \frac{1}{180}\tilde{F}_{j_2} & \frac{1}{6}\tilde{F}_{j_1} & \tilde{F}_{j_0} + \frac{1}{3}\tilde{F}_{j_2} & \tilde{F}_{j_1} & \tilde{F}_{j_2} & 0 & 0 & 0 \\ \frac{-1}{120}\tilde{F}_{j_1} & 0 & \frac{1}{4}\tilde{F}_{j_1} & \tilde{F}_{j_0} + \frac{1}{2}\tilde{F}_{j_2} & \tilde{F}_{j_1} & \tilde{F}_{j_2} & 0 & 0 \\ \frac{-1}{630}\tilde{F}_{j_2} & \frac{-1}{30}\tilde{F}_{j_1} & \frac{-1}{30}\tilde{F}_{j_2} & \frac{1}{3}\tilde{F}_{j_1} & \tilde{F}_{j_0} + \frac{2}{3}\tilde{F}_{j_2} & \tilde{F}_{j_1} & \tilde{F}_{j_2} & 0 \\ \frac{1}{252}\tilde{F}_{j_1} & 0 & \frac{-1}{12}\tilde{F}_{j_1} & \frac{-1}{2}\tilde{F}_{j_2} & \frac{5}{12}\tilde{F}_{j_1} & \frac{5}{6}\tilde{F}_{j_2} + \tilde{F}_{j_0} & \tilde{F}_{j_1} & \tilde{F}_{j_2} \\ \frac{1}{840}\tilde{F}_{j_2} & \frac{1}{42}\tilde{F}_{j_1} & \frac{1}{42}\tilde{F}_{j_2} & \frac{-1}{6}\tilde{F}_{j_1} & \frac{-1}{4}\tilde{F}_{j_2} & \frac{1}{2}\tilde{F}_{j_1} & \tilde{F}_{j_0} + \tilde{F}_{j_2} & \tilde{F}_{j_1} \\ \frac{-1}{240}\tilde{F}_{j_1} & 0 & \frac{1}{12}\tilde{F}_{j_1} & \frac{1}{9}\tilde{F}_{j_2} & \frac{-7}{24}\tilde{F}_{j_1} & \frac{-7}{15}\tilde{F}_{j_2} & \frac{7}{12}\tilde{F}_{j_1} & \tilde{F}_{j_0} + \frac{7}{6}\tilde{F}_{j_2} \end{bmatrix}, \tag{19}$$

where 0 and  $\tilde{F}_{ji}$ ,  $i = 0, 1, 2$ , are the  $3 \times 3$  matrices given by

$$\tilde{F}_{ji} = \begin{bmatrix} f_{1i}^j & 0 & 0 \\ 0 & f_{2i}^j & 0 \\ 0 & 0 & f_{3i}^j \end{bmatrix}.$$

By integrating Equation (12) from 0 to  $t$  and using Equations (15)-(22), we have

$$\hat{\phi}^T(t) X - \hat{\phi}^T(t) d = \sum_{j=0}^r \hat{\phi}^T(t) \hat{P}^T \tilde{F}_j^T R_j + \hat{\phi}^T(t) Z_j \tilde{F}_j^T R_j + \hat{\phi}^T(t) \hat{P}^T \tilde{F}_j^T \hat{D}_j^T X + \hat{\phi}^T(t) \hat{P}^T \tilde{G}^T U, \quad (20)$$

simplifying Equation (23) we obtain

$$\left[ I - \sum_{j=0}^r \hat{P}^T \tilde{F}_j^T \hat{D}_j^T \right] X = d + \hat{P}^T \tilde{G}^T U + \sum_{j=0}^r \hat{P}^T \tilde{F}_j^T R_j + Z_j \tilde{F}_j^T R_j. \quad (21)$$

by solving the set of linear algebraic equations Equation (24), we obtain the coefficients vector  $X$ .

### 4. Numerical Implementation

In this section, to give a clear overview of the analysis method presented and to demonstrate the applicability and accuracy of the method three examples are given.

**Example 1.** Consider the multi-delay dynamic system from [7] described by

$$\begin{pmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{pmatrix} = \begin{pmatrix} t & 1 \\ t & 2t \end{pmatrix} \begin{pmatrix} x_1\left(t - \frac{1}{3}\right) \\ x_2\left(t - \frac{1}{3}\right) \end{pmatrix} + \begin{pmatrix} 2 & t \\ t^2 & 0 \end{pmatrix} \begin{pmatrix} x_1\left(t - \frac{2}{3}\right) \\ x_2\left(t - \frac{2}{3}\right) \end{pmatrix} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u(t), \quad (22)$$

with

$$x_1(t) = x_2(t) = u(t) = 0, \quad t \in \left[-\frac{2}{3}, 0\right], \quad (23)$$

and

$$u(t) = 2t + 1, \quad t > 0. \quad (24)$$

The exact solutions are

$$x_1(t) = \begin{cases} 0 & 0 \leq t \leq \frac{1}{3}, \\ \frac{7}{162} - \frac{2}{9}t + \frac{1}{6}t^2 + \frac{1}{3}t^3 & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ \frac{11}{162} - \frac{58}{243}t + \frac{31}{162}t^2 + \frac{1}{9}t^3 + \frac{7}{72}t^4 + \frac{1}{6}t^5 & \frac{2}{3} \leq t \leq 1. \end{cases}$$

$$x_2(t) = \begin{cases} t + t^2 & 0 \leq t \leq \frac{1}{3}, \\ \frac{5}{486} + t + \frac{7}{9}t^2 + \frac{2}{9}t^3 + \frac{1}{2}t^4 & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ \frac{1}{486} + t + \frac{200}{243}t^2 + \frac{20}{81}t^3 + \frac{29}{72}t^4 - \frac{1}{9}t^5 + \frac{1}{6}t^6 & \frac{2}{3} \leq t \leq 1. \end{cases}$$

To solve this problem by the hybrid functions, we select  $N = 3$  and  $M = 7$ . Let

$$x_1(t) = X_1^T \phi(t), \quad x_2(t) = X_2^T \phi(t), \quad (25)$$

where  $X_1^T$ ,  $X_2^T$  and  $\phi(t)$  can be obtained similarly to Equations (3) and (4). By expanding  $t$  and  $t^2$  in terms of the hybrid functions we get

$$t = \left[ \frac{1}{6}, \frac{1}{2}, \frac{5}{6}, \frac{1}{3}, \frac{1}{3}, \frac{1}{3}, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0 \right] \phi(t) = F_1^T \phi(t), \quad (26)$$

$$t^2 = \left[ \frac{1}{27}, \frac{7}{27}, \frac{19}{27}, \frac{1}{9}, \frac{1}{3}, \frac{5}{9}, \frac{1}{9}, \frac{1}{9}, \frac{1}{9}, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0 \right] \phi(t) = F_2^T \phi(t), \tag{27}$$

Therefore, we have

$$tx_1 \left( t - \frac{1}{3} \right) = X_1^T D_1 \tilde{F}_1 \phi(t), \quad tx_2 \left( t - \frac{1}{3} \right) = X_2^T D_1 \tilde{F}_1 \phi(t), \tag{28}$$

$$t^2 x_1 \left( t - \frac{2}{3} \right) = X_1^T D_2 \tilde{F}_2 \phi(t), \quad tx_2 \left( t - \frac{2}{3} \right) = X_2^T D_2 \tilde{F}_1 \phi(t), \tag{29}$$

where  $\tilde{F}_1$  and  $\tilde{F}_2$  the  $24 \times 24$  matrices, can be calculated as Equation (19). Also  $D_1$  and  $D_2$  are the  $24 \times 24$  delay operational matrices given by

$$D_1 = \text{diag}(\psi_1, \psi_1, \dots, \psi_1), \quad D_2 = \text{diag}(\psi_2, \psi_2, \dots, \psi_2),$$

where

$$\psi_1 = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix}, \quad \psi_2 = \begin{bmatrix} 0 & 0 & 1 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}.$$

Integrating Equation (25) from 0 to  $t$  and using Equations (26)-(27) and substituting Equations (28)-(32) we get

$$\begin{cases} X_1^T (P - D_1 \tilde{F}_1 P - 2D_2 P) - X_2^T D_1 P = 0, \\ X_1^T (-D_1 \tilde{F}_1 P - D_2 \tilde{F}_2 P) + X_2^T (P - 2D_1 \tilde{F}_1 P) = F_1 + F_2, \end{cases} \tag{30}$$

where  $P$  is the operational matrix of integration given in Equation (7). By solving Equation (33) the values of  $X_1^T$  and  $X_2^T$  can be found as

$$X_1^T = \left[ 0, \frac{7}{324}, \frac{17251}{87480}, 0, \frac{11}{162}, \frac{637}{1944}, 0, \frac{2}{27}, \frac{305}{1458}, 0, \frac{1}{81}, \frac{521}{8748}, 0, 0, \frac{19}{1944}, 0, 0, \frac{1}{1458}, 0, 0, 0, 0, 0, 0 \right],$$

$$X_2^T = \left[ \frac{11}{54}, \frac{949}{1215}, \frac{3287617}{1837080}, \frac{4}{9}, \frac{361}{486}, \frac{2611}{1944}, \frac{1}{9}, \frac{17}{81}, \frac{1223}{2916}, 0, \frac{11}{243}, \frac{2735}{26244}, 0, \frac{1}{162}, \frac{367}{17496}, 0, 0, \frac{13}{4374}, 0, 0, \frac{1}{4374}, 0, 0, 0 \right].$$

To define  $x_1(t)$  and  $x_2(t)$  for  $t$  in the interval  $\left[ 0, \frac{1}{3} \right]$  we map  $\left[ 0, \frac{1}{3} \right]$  into  $[0,1]$  by mapping  $t$  into  $3t$ , and for  $t$  in the interval  $\left[ \frac{1}{3}, \frac{2}{3} \right]$  we map this interval into  $[0,1]$  by mapping  $t$  into  $3t-1$ , and similarly for the other intervals. From Equation (28) we get

$$x_1(t) = \begin{cases} c_{10} \beta_0(3t) + c_{11} \beta_1(3t) + c_{12} \beta_2(3t) + c_{13} \beta_3(3t) + c_{14} \beta_4(3t) + c_{15} \beta_5(3t) \\ \quad + c_{16} \beta_6(3t) + c_{17} \beta_7(3t) & 0 \leq t \leq \frac{1}{3}, \\ c_{120} \beta_0(3t-1) + c_{121} \beta_1(3t-1) + c_{122} \beta_2(3t-1) + c_{123} \beta_3(3t-1) + c_{124} \beta_4(3t-1) \\ \quad + c_{125} \beta_5(3t-1) + c_{126} \beta_6(3t-1) + c_{127} \beta_7(3t-1) & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ c_{130} \beta_0(3t-2) + c_{131} \beta_1(3t-2) + c_{132} \beta_2(3t-2) + c_{133} \beta_3(3t-2) + c_{134} \beta_4(3t-2) \\ \quad + c_{135} \beta_5(3t-2) + c_{136} \beta_6(3t-2) + c_{137} \beta_7(3t-2) & \frac{2}{3} \leq t \leq 1. \end{cases}$$

$$x_2(t) = \begin{cases} c_{2_{10}}\beta_0(3t) + c_{2_{11}}\beta_1(3t) + c_{2_{12}}\beta_2(3t) + c_{2_{13}}\beta_3(3t) + c_{2_{14}}\beta_4(3t) + c_{2_{15}}\beta_5(3t) \\ \quad + c_{2_{16}}\beta_6(3t) + c_{2_{17}}\beta_7(3t) & 0 \leq t \leq \frac{1}{3}, \\ c_{2_{20}}\beta_0(3t-1) + c_{2_{21}}\beta_1(3t-1) + c_{2_{22}}\beta_2(3t-1) + c_{2_{23}}\beta_3(3t-1) + c_{2_{24}}\beta_4(3t-1) \\ \quad + c_{2_{25}}\beta_5(3t-1) + c_{2_{26}}\beta_6(3t-1) + c_{2_{27}}\beta_7(3t-1) & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ c_{2_{30}}\beta_0(3t-2) + c_{2_{31}}\beta_1(3t-2) + c_{2_{32}}\beta_2(3t-2) + c_{2_{33}}\beta_3(3t-2) + c_{2_{34}}\beta_4(3t-2) \\ \quad + c_{2_{35}}\beta_5(3t-2) + c_{2_{36}}\beta_6(3t-2) + c_{2_{37}}\beta_7(3t-2) & \frac{2}{3} \leq t \leq 1. \end{cases}$$

After simplifying the same value as the exact  $x_1(t)$  and  $x_2(t)$  would be obtained.

**Example 2.** Consider the delay dynamic system described by

$$\begin{cases} \dot{x}_1(t) = x_2(t) \\ \dot{x}_2(t) = x_2\left(t - \frac{1}{2}\right) + u(t) \end{cases} \quad 0 \leq t \leq 1 \tag{31}$$

with

$$x_1(t) = x_2(t) = u(t) = 0, \quad t \in \left[-\frac{1}{2}, 0\right], \tag{32}$$

and

$$u(t) = t, \quad t > 0. \tag{33}$$

The exact solutions are

$$x_1(t) = \begin{cases} \frac{1}{6}t^3 & 0 \leq t \leq \frac{1}{2}, \\ \frac{1}{48} + \frac{\left(t - \frac{1}{2}\right)}{8} + \frac{\left(t - \frac{1}{2}\right)^2}{4} + \frac{\left(t - \frac{1}{2}\right)^3}{6} + \frac{\left(t - \frac{1}{2}\right)^4}{24} & \frac{1}{2} \leq t \leq 1. \end{cases}$$

$$x_2(t) = \begin{cases} \frac{t^2}{2} & 0 \leq t \leq \frac{1}{2}, \\ \frac{1}{8} + \frac{\left(t - \frac{1}{2}\right)}{2} + \frac{\left(t - \frac{1}{2}\right)^2}{2} + \frac{\left(t - \frac{1}{2}\right)^3}{6} & \frac{1}{2} \leq t \leq 1. \end{cases}$$

To solve this problem by using of the hybrid functions, we select  $N = 2$  and  $M = 5$ . Let

$$x_1(t) = X_1^T \phi(t), \quad x_2(t) = X_2^T \phi(t), \tag{34}$$

where  $X_1^T$ ,  $X_2^T$  and  $\phi(t)$  can be obtained similarly to Equations (3)-(4). Using Equation (37) we get

$$x_2\left(t - \frac{1}{2}\right) = X_2^T D_1 \phi(t),$$

where  $D_1$  is the  $12 \times 12$  delay operational matrix given by

$$D_1 = \text{diag}(\psi_1, \psi_1, \dots, \psi_1),$$

where

$$\psi_1 = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}, \tag{35}$$



By expanding  $u(t)$  in terms of hybrid functions we obtain

$$u(t) = U^T \phi(t). \tag{36}$$

Integrating Equation (34) from 0 to  $t$  and using Equations (35)-(36) and substituting Equations (37)-(39), we get

$$\begin{cases} X_1^T - X_2^T P = 0, \\ X_2^T (I - D_1 P) = U^T P, \end{cases} \tag{37}$$

where  $P$  is the operational matrix of integration given in Equation (7). By solving Equation (40) the values of  $X_1^T$  and  $X_2^T$  can be found. By using from Equation (37) and simplifying the same value as the exact  $x_1(t)$  and  $x_2(t)$  would be obtained.

**Example 3.** Consider the following multi-delay system with delay in both control and state described by

$$\begin{cases} \dot{x}(t) = x_2(t) \\ \dot{x}_2(t) = x_1(t) + x_2\left(t - \frac{1}{3}\right) + u\left(t - \frac{2}{3}\right) \end{cases} \quad 0 \leq t \leq 1 \tag{38}$$

with

$$x_1(t) = x_2(t) = u(t) = 0, \quad t \in \left[-\frac{2}{3}, 0\right), \quad u(t) = 1, \quad t > 0, \tag{39}$$

and

$$x_1(0) = x_2(0) = 0. \tag{40}$$

The exact solutions are

$$x_1(t) = \begin{cases} \cosh(t) & 0 \leq t \leq \frac{1}{3}, \\ \cosh(t) + \frac{t}{2} \cosh\left(t - \frac{1}{3}\right) + \frac{1}{6} e^{-\left(t - \frac{1}{3}\right)} - \frac{1}{3} e^{-t} & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ -\frac{1}{4} \sinh(t-1) + \frac{t}{4} \cosh\left(\frac{1}{3} - t\right) + \frac{1}{16} t^2 \cosh\left(t - \frac{2}{3}\right) + \cosh\left(\frac{2}{3}\right) \cosh\left(t - \frac{2}{3}\right) \\ + \frac{1}{3} \sinh\left(\frac{1}{3}\right) \sinh\left(t - \frac{2}{3}\right) + \sinh\left(\frac{1}{3}\right) \sinh\left(t - \frac{2}{3}\right) + \frac{1}{6} \sinh\left(t - \frac{2}{3}\right) \cosh\left(\frac{1}{3}\right) \\ - \frac{t}{24} \cosh\left(t - \frac{2}{3}\right) + \frac{35}{36} \cosh\left(t - \frac{2}{3}\right) - \frac{1}{4} \sinh\left(t - \frac{1}{3}\right) - \frac{1}{3} e^{-\frac{t-1}{3}} + \left(\frac{1}{12} - \frac{t}{8}\right) e^{-\frac{t-2}{3}} + \frac{1}{6} e^{t-1} - 1 & \frac{2}{3} \leq t \leq 1 \end{cases}$$

$$x_2(t) = \begin{cases} \sinh(t) & 0 \leq t \leq \frac{1}{3}, \\ \sinh(t) - \frac{t}{2} \sinh\left(\frac{1}{3} - t\right) + \frac{1}{2} \cosh\left(\frac{1}{3} - t\right) - \frac{1}{3} e^{-\frac{t-1}{3}} - \frac{1}{6} e^{-t} & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ -\frac{1}{4} \cosh(t-1) + \frac{t}{2} \sinh\left(t - \frac{1}{3}\right) + \frac{1}{8} t^2 \sinh\left(t - \frac{2}{3}\right) + \cosh\left(\frac{2}{3}\right) \sinh\left(t - \frac{2}{3}\right) \\ + \frac{1}{6} \cosh\left(\frac{1}{3}\right) \cosh\left(t - \frac{2}{3}\right) + \sinh\left(\frac{2}{3}\right) \cosh\left(t - \frac{2}{3}\right) + \frac{1}{3} \cosh\left(t - \frac{2}{3}\right) \sinh\left(\frac{1}{3}\right) \\ - \frac{t}{24} \sinh\left(t - \frac{2}{3}\right) + \frac{61}{72} \sinh\left(t - \frac{2}{3}\right) - \frac{1}{4} \sinh\left(t - \frac{1}{3}\right) - \frac{1}{12} e^{-\frac{t-1}{3}} + \left(\frac{t}{8} - \frac{1}{12}\right) e^{-\frac{t-2}{3}} + \frac{1}{6} e^{t-1} & \frac{2}{3} \leq t \leq 1. \end{cases}$$

To solve this problem by using the hybrid functions, we select  $N = 3$  and  $M = 7$ . Let

$$x_1(t) = X_1^T \phi(t), \quad x_2(t) = X_2^T \phi(t), \tag{41}$$

where  $X_1^T$ ,  $X_2^T$  and  $\phi(t)$  can be obtained similarly to Equations (3) and (4). By expanding  $u(t)$  and  $x_1(0)$  in terms of hybrid functions we get

$$x_1(0) = E^T \phi(t), \quad u(t) = E^T \phi(t). \tag{42}$$

Using Equation (44) and (45) we obtain

$$\begin{aligned} u\left(t - \frac{2}{3}\right) &= E^T D_2 \phi(t), \\ x_2\left(t - \frac{2}{3}\right) &= X_2^T D_1 \phi(t), \end{aligned} \tag{43}$$

where  $D_1$  and  $D_2$  are the  $24 \times 24$  delay operational matrices given by

$$D_1 = \text{diag}(\psi_1, \psi_1, \dots, \psi_1), \quad D_2 = \text{diag}(\psi_2, \psi_2, \dots, \psi_2), \tag{44}$$

where

$$\psi_1 = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix}, \quad \psi_2 = \begin{bmatrix} 0 & 0 & 1 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}.$$

By integrating Equation (41) from 0 to  $t$  and using Equations (42) and (43) and substituting Equations (44) and (47), we get

$$\begin{cases} X_1^T - X_2^T P = E^T, \\ -X_1^T P + X_2^T (I - D_1 P) = E^T D_2 P, \end{cases} \tag{45}$$

where  $P$  is the operational matrix of integration given in Equation (7). By solving Equation (48) the values of  $X_1^T$  and  $X_2^T$  can be found. By using from Equation (44) and simplifying the same value as the exact  $x_1(t)$  and  $x_2(t)$  would be obtained. In **Table 1** a comparison is made between the exact solution and the approximation solution of  $x_1(t)$  and  $x_2(t)$  for  $0 \leq t \leq 1$ . The approximation value of  $x_1(t)$  and  $x_2(t)$  on  $[0, 1]$ , is the same as the exact solution.

$$\begin{aligned} x_1(t) &= \begin{cases} 1. + 2.1196 \times 10^{-11} t + 0.5t^2 - 1.26823 \times 10^{-9} t^3 + 0.0416674t^4 \\ \quad + 2.39729 \times 10^{-8} t^5 + 0.00137581t^6 + 0.0000333761t^7 & 0 \leq t \leq \frac{1}{3}, \\ 0.993758 + 0.0565911t + 0.3271t^2 + 0.185487t^3 + 0.0129901t^4 \\ \quad + 0.0182755t^5 - 0.000222216t^6 + 0.000713909t^7 & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ 1.2329 - 0.712896t + 1.06433t^2 - 0.066174t^3 + 0.128371t^4 \\ \quad + 0.000320482t^5 + 0.00331394t^6 + 0.00112221t^7 & \frac{2}{3} \leq t \leq 1 \end{cases} \\ x_2(t) &= \begin{cases} 2.1196 \times 10^{-11} + t - 3.80469 \times 10^{-9} t^2 + 0.166667t^3 + 1.19864 \times 10^{-7} t^4 \\ \quad + 0.00833348t^5 - 2.15927 \times 10^{-6} t^6 + 0.000202107t^7 & 0 \leq t \leq \frac{1}{3}, \\ 0.05659 + 0.654221t + 0.556305t^2 + 0.0525749t^3 + 0.0899829t^4 \\ \quad + 0.000479554t^5 + 0.00374231t^6 + 0.000358585t^7 & \frac{1}{3} \leq t \leq \frac{2}{3}, \\ -0.713338 + 2.1327t - 0.214113t^2 + 0.546471t^3 - 0.0396684t^4 \\ \quad + 0.0504241t^5 - 0.00452581t^6 + 0.0021225t^7 & \frac{2}{3} \leq t \leq 1 \end{cases} \end{aligned}$$

**Table 1.** Approximate solutions and exact solutions of Example 3.

$t$	$x_1(t)$		$x_2(t)$	
	Approximatesolution	Exactsolution	Approximatesolution	Exactsolution
0	1	1	$2.1196 \times 10^{-11}$	0
0.1	1.00500	1.00500	0.100167	0.100167
0.2	1.02007	1.02007	0.201336	0.201336
0.3	1.04534	1.04534	0.304520	0.304520
0.4	1.08112	1.08112	0.412976	0.412976
0.5	1.12840	1.12840	0.535049	0.535049
0.6	1.18865	1.18865	0.672632	0.672632
0.7	1.26405	1.26405	0.860668	0.860668
0.8	1.36366	1.36366	1.135120	1.135120
0.9	1.49187	1.49187	1.433400	1.433400
1	1.65129	1.65129	1.760070	1.760070

## 5. Conclusion

The hybrid of the Block-Pulse functions and the Bernoulli polynomials and the associated operational matrices of integration and delay are applied to solve the linear multi-delay dynamic systems. The method is computationally very attractive, at the same time keeping the accuracy of the solution. It is also shown that the hybrid functions provide exact solutions in each subintervals for Examples 1, 2 and 3. The presented method reduces multi-delay systems to the solution a system of algebraic equations, and so the calculation is easy. The matrices  $K$ ,  $P$  and  $D_j$  in Equations (5), (7) and (10) are sparse, hence the present method is very attractive and reduces the CPU time and computer memory.

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