

A Novel Methodology to Design Miniaturized Regular Planar Inverted-F Antennas Based on Parametric Simulations

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ABSTRACT

This paper describes a novel methodology to design PIFA (Planar Inverted-F Antenna) antennas based on parametric simulations. The parameters considered in the parametric design are ground plan dimensions, height of radiating plate, feeding point position, shorting plate width and position. The choice of the parameters that must be varied independently or simultaneously is important to design optimized antenna. The author studied two scenarios in precedent works [1,2]. He exposes here a third scenario of varying antenna parameters to design and simulate by HFSS (High Frequency Simulator Structure) simulator a probe-fed dual band PIFA for the use in GSM 850 band (824 MHz - 894 MHz) and PCS 1900 band (1850 MHz - 1990 MHz). The author compares the three scenarios and establishes a novel methodology to design optimized and miniaturized antennas mounted on mobile handsets.

Keywords: PIFA; HFSS; Design; Methodology

1. Introduction

With the development of mobile networks and wireless applications, the handset should be more miniature, while working in different bands and supporting different applications. A small mobile device currently supports GSM services, high speed mobile internet access and allows connection via WiFi and blue tooth capabilities. For this purpose, antennas are vowed to achieve miniaturization while maintaining function requirements. The planar antennas are so compliant and their use in wireless local and wide networks increases.

For optimum system performance, the antennas must have high radiation efficiency, small volume, isotropic radiation characteristics, small backward radiation, simple and low-loss impedance matching to patches. The major types of configurations of low-profile antennas with enhanced bandwidth performance include Planar Inverted-F Antennas.

The PIFA consists in general of a ground plane, a top plate element, a feed wire attached between the ground plane and the top plate, and a shorting wire or strip that is connected between the ground plane and the top plate.

The antenna is fed at the base of the feed wire at the

point where the wire connects to the ground plane. The PIFA is an attractive antenna for wireless systems where the space volume is quite limited. It requires simple manufacturing, since the radiator must only be printed. The addition of a shorting strip allows good impedance match to be achieved with a top plate that is typically less than $\lambda/4$ long. The resulting PIFA is more compact than a conventional half-wavelength probe-fed patch antenna [3].

The miniaturization can affect radiation characteristics, bandwidth, gain, radiation efficiency and polarization purity. The miniaturization approaches are based on either geometric manipulation (the use of bend forms, meandered lines, PIFA shape, varying distance between feeder and short plate [4]) or material manipulation (loading with a high-dielectric material, lumped elements, conductors, capacitors, short plate [5]) or the environment characteristics (ground plane dimensions, coupling, measurement and fabrication errors [4]). In this case, the bi-band designed antenna is shorted to the ground plane by a plate, uses regular shapes and uses a high dielectric thin substrate under the radiating plate not above the ground plane).

If all precedent works are concentrated on studying the effects of these elements (material, geometry, environment), the choice of a PIFA element was so improvised in the design. There are some recent works to make algorithms or parametric simulations but they use some models concerning the traditional patch antennas [6] or they deliver the theoretical frequency from predefined parameters [7,8]. The author worked on parametric simulations by varying sequently the antenna parameters (scenario 1) [1] or simultaneously of some parameters (scenario 2) [2]. In this paper, the author will expose a third scenario and he will compare the three scenarios before concluding about efficient methodology to design PIFA single or multiband.

The third scenario will be applied to design a dual band PIFA for the use in the GSM850 and PCS1900 bands.

2. Antenna Design Following the Third Scenario

2.1. The Description of the Studied Antenna

As shown in **Figure 1**, the designed antenna has a rectangular radiating patch length $L_p = 31$ mm and width $W_p = 70$ mm. The patch is placed at a height h from the ground plan. The ground plan has a length L_g and a width W_g equal to W_p . The patch is matched to the ground plan via a rectangular shorting plate. The shorting plate has a width W_s and a length and it is placed in the (yz) plan at a distance D from the edge centre. The the feeding point is situated at p from the rear edge of the patch. The patch is fed by a 50Ω wire. The volume between the radiating plate and the ground plan is filled by air except a thin region 0.8 mm under the radiating patch who is composed of FR4_epoxy ($\epsilon_r = 4.4$).

2.2. The Choice of the Patch Dimensions

It is very important for simulation by HFSS to estimate the resonant frequency that help the simulator to make a refinement mesh in a band around the resonant frequency and then give more precise values. The resonant frequency of a PIFA is approximated by the Equation (1) [9] where F_r is the resonant frequency, C is the light velocity.

$$F_r = \frac{C}{4(L_p + W_p - W_s)} \quad (1)$$

If there is another substrate different from Air, C will be $C_0/\sqrt{\epsilon_r}$ where $C_0 = 3 \times 10^8$ m/s. For our case, the space between the patch and the ground plan is essentially air minus a 0.8 mm FR4 epoxy layer.

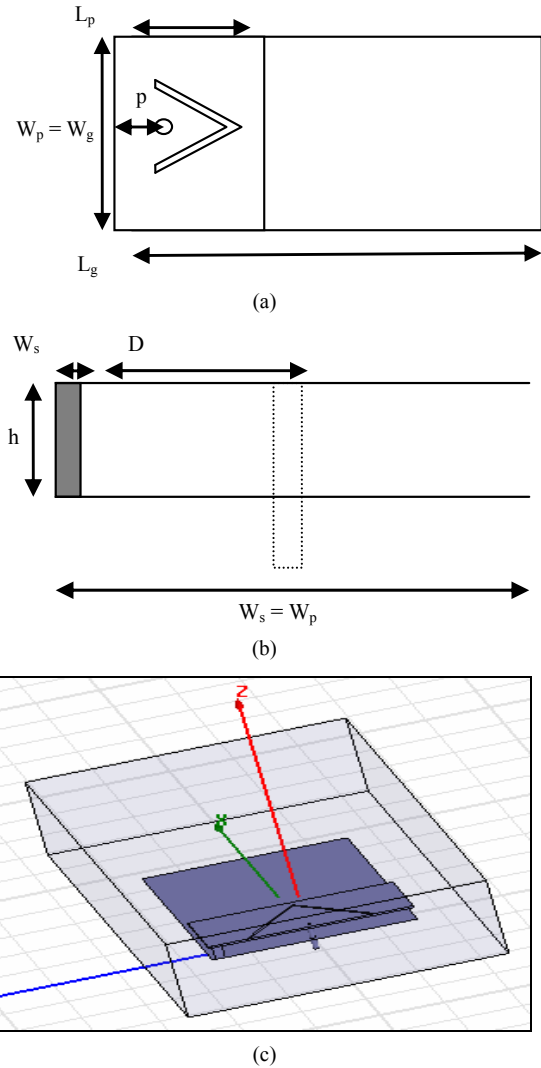


Figure 1. Different views of the designed antenna; (a) front view; (b) side view; (c) perspective view.

To compute the resonant frequency, we have the following values: $L_p = 31$ mm, $W_p = 70$ mm, $W_{smax} = 3$ mm. The theoretical Fris 790 MHz. The obtained frequency is then not far from 824 MHz the first frequency of GSM850 band. In fact, there is no equation (not empirical) to determine the resonant frequency for a PIFA that contains not only the patch dimensions but also the other parameters that can affect the antenna characteristics. For this, the author will make constant the patch dimensions that are the mean parameters can furnish the resonant frequency and he will vary (following the scenario 3) firstly the height, secondly and simultaneously the length L_g , the width W_s , the distance D . Finally, the author will vary the feeding point position p and at last and simultaneously undependably the other parameters (ground plan dimensions L_g , height of radiating plate h , feeding point position p , shorting plate Width W_s and position D).

2.3. The Choice of the Height h

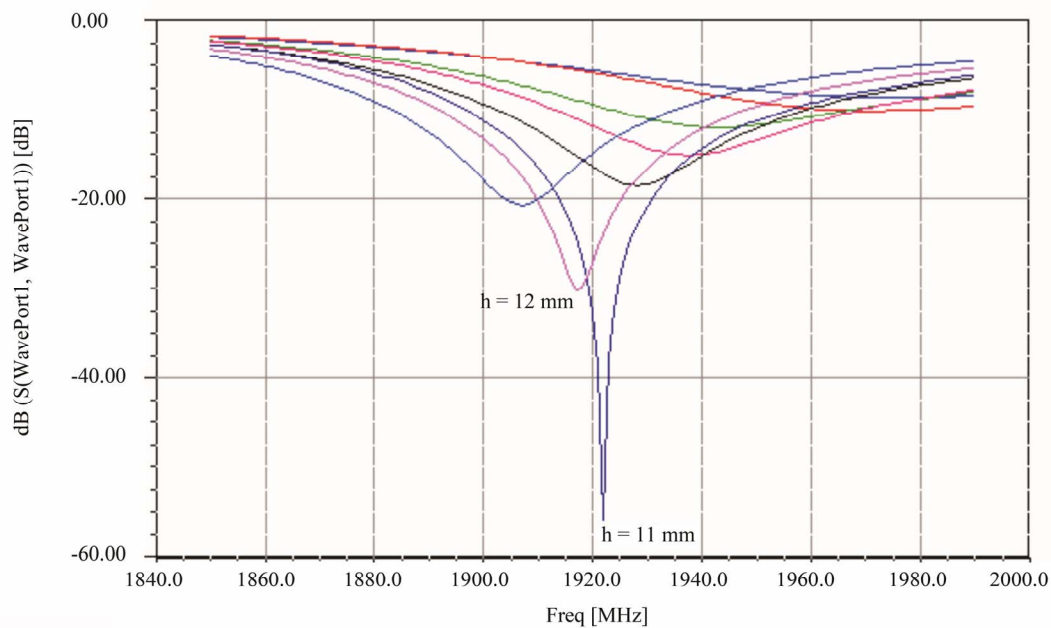
The height h is the distance between the top plate and the ground plane. In order to eliminate the effects of the ground plane, the patched is placed on the edge of an infinite ground plane (in HFSS, the choice of the infinite ground plan can be made during the definition of boundaries) at a height varying from 6 mm to 13 mm.

From the simulation result shown by **Figure 2**, the optimal height h is for $h = 11$ mm because it gives values S_{11} most important and closer to central frequencies of

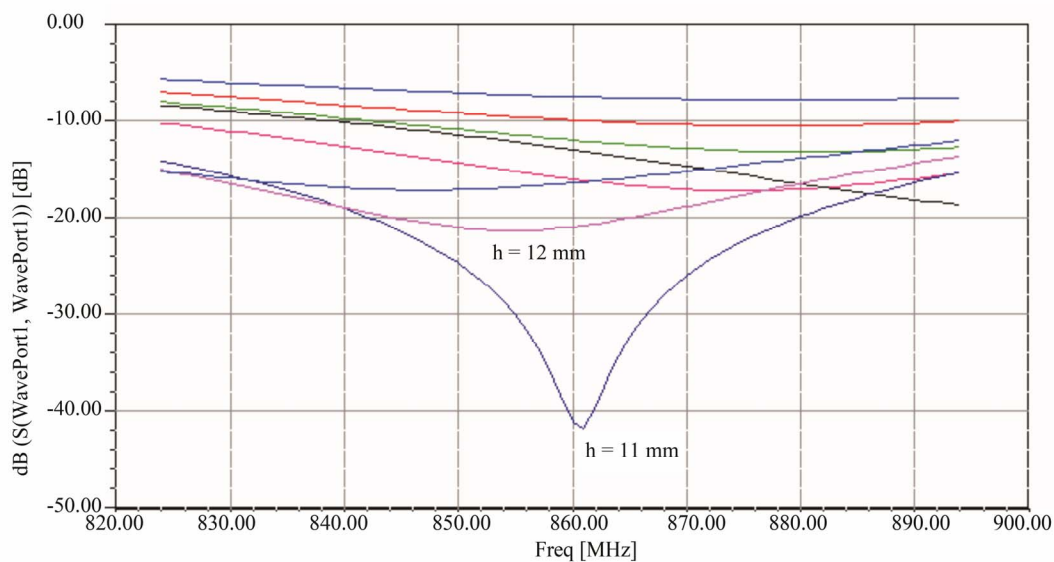
both bands (859 MHz and 1920 MHz). This height is a compliant (it will be nearly the handset thickness).

2.4. The simultaneous Choice of the Ground Plan Length L_g , the Short Plate Position D and the Width W_s

The height h is taken then equal to 11 mm. It's an adequate height because the PIFA will be mounted on a GSM handset (the height is practically close to handset thickness). We will vary simultaneously the ground plane



(a)



(b)

Figure 2. h -parametric simulation results for GSM850 and PCS bands. (a) for GSM850; (b) for PCS1900.

length L_g from 80 mm to 140 mm, D will vary from 29 mm to 34 mm and the width W_s from 1 mm to 3 mm. We

will then choose our antenna configuration from $7 \times 6 \times 3 = 126$ possibilities. The **Figures 3(a)** shows the results

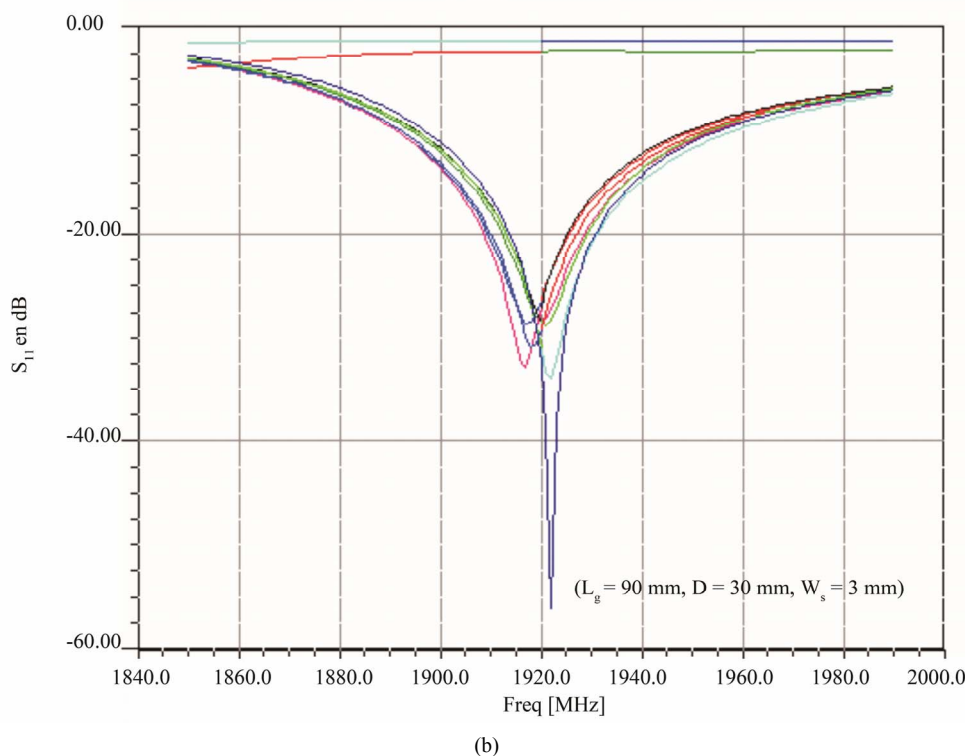
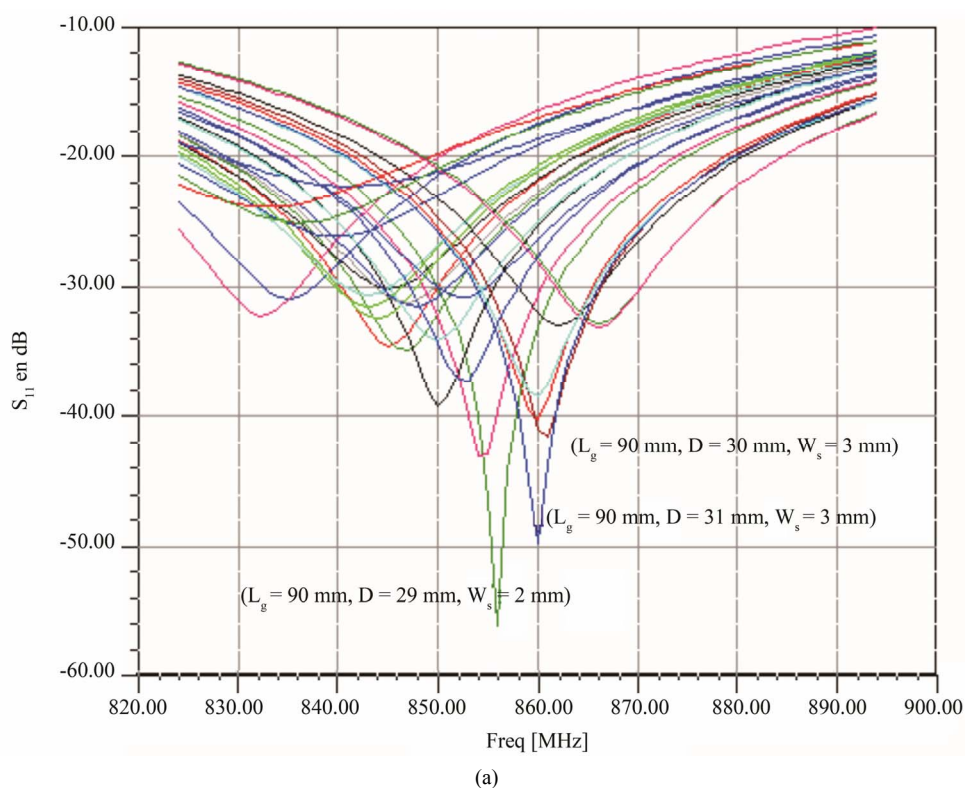


Figure 3. Tri-parametric simulation results; (a) for GSM850 band; (b) for PCS1900 band.

tion ($L_g = 90$ mm, $D = 30$ mm, $W_s = 3$ mm) that is one of the three configurations related in **Figure 3**. We can of the tri-parametric simulation for GSM 850 band. From **Figure 3**, there are three curves that present minimal S_{11} values and close to the central frequency 859 MHz. The three configurations are as indicated in **Figure 3(a)** ($L_g = 90$ mm, $D = 31$ mm, $W_s = 3$ mm), ($L_g = 90$ mm, $D = 30$ mm, $W_s = 3$ mm) and ($L_g = 90$ mm, $D = 29$ mm, $W_s = 2$ mm). We Will now check the simulation result for the three configurations but for PCS band. The result is shown in **Figure 3(b)**.

The **Figure 3(b)** shows a S_{11} peak for the configura- consider the triplet ($L_g = 90$ mm, $D = 30$ mm, $W_s = 3$ mm) a very interesting trade-off configuration for both bands.

2.5. The Choice of the Feeding Point Position p

We have then chosen the parameters h , L_g , D and W_s . We can now look for the effect of the feeding point position p to enhance the bandwidth or the input impedance. It is calculated from the rear edge of the patch. We will vary the p parameter from 2 mm to 16 mm (the centre). The value $p = 1$ can't be taken because the feeding point is theoretically a circle that has a radius. The result is shown on **Figure 4**. We can consider the peak for $p = 3$ mm an interesting position for feeding for PCS band. We will see for GSM band as shown in **Figure 5**. It is an adequate position because it gives a S_{11} peak very closely to the central frequency of the GSM850 band.

2.6 The Third Scenario Results

The results of parametric simulations are exposed in the following. We can see in **Figure 6** two peaks of S_{11} parameter, one is around 860 MHz (very close to the GSM850 central frequency 859 MHz), the second peak is around 1922 MHz (very close to the PCS central frequency 1920 MHz). Also, the S_{11} values out of both bands are near 0, that means the designed antenna can't interfere with other radiations. We can also run the simulation by refining the sweep interval for more precision. As given exactly by simulations tables, we note $S_{11} = -14.04$ dB for 824 MHz (the low frequency of the GSM850 band), $S_{11} = -14.89$ dB for 894 MHz (the high frequency of the GSM band), $S_{11} = -2.7$ dB for 1850 MHz (the low frequency of the PCS band), $S_{11} = -6.09$ dB for 1990 MHz (the high frequency of the PCS band).

2.6.1. The Bandwidth and VSWR

We obtain as shown in **Figure 7(a)** for GSM850 band a VSWR = 1.49 for 824 MHz (the lowest frequency), VSWR min = 1.03 for 860 MHz (the resonant frequency), VSWR = 1.43 for 894 MHz (the highest frequency). The VSWR is at its minimum, it's very interesting result. Also, The GSM bandwidth (70 MHz) is for the designed antenna a 1:1.5 VSWR bandwidth and the antenna presents a presents a 1:2 bandwidth equal to 120 MHz. It is considered a very interesting result.

Also, we obtain as shown in **Figure 7(b)** for PCS band

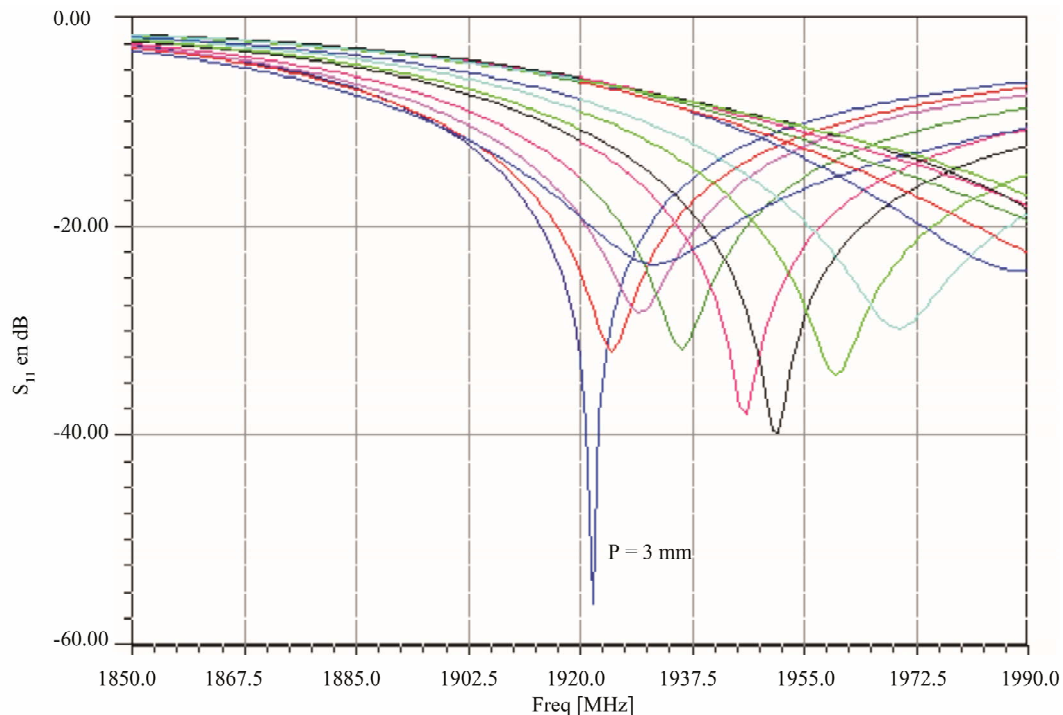


Figure 4. Feeding point position parametric simulation results for PCS1900 band.

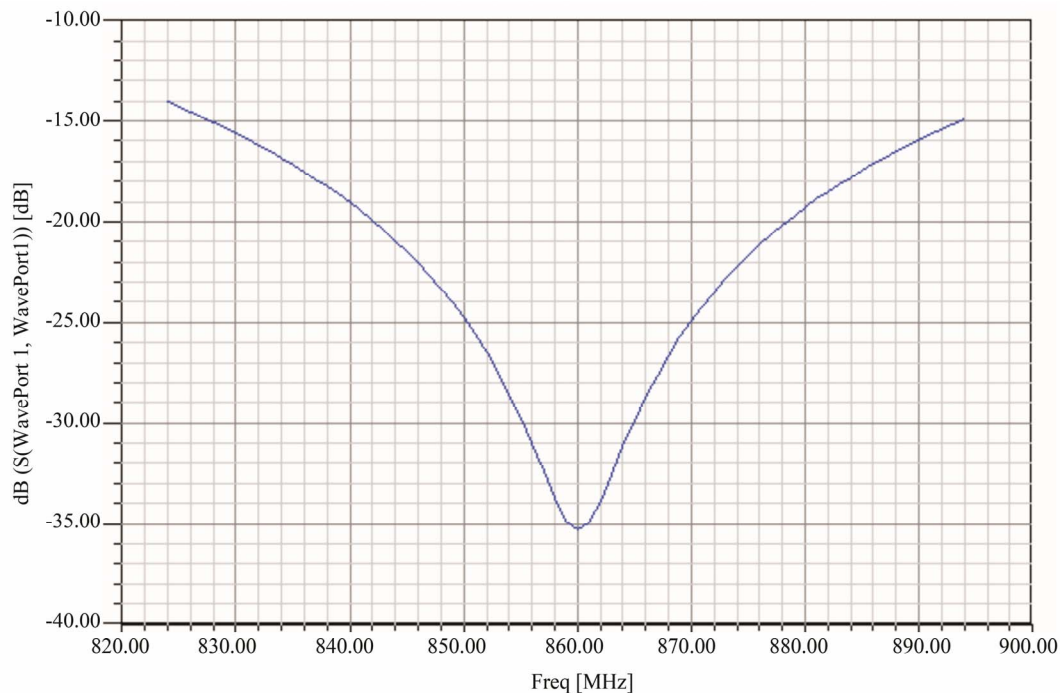


Figure 5. The selected configuration in GSM850 band.

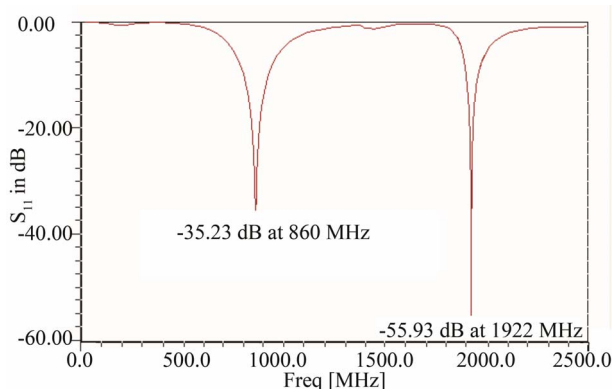


Figure 6. S_{11} depending on frequency for (0 - 2400 MHz).

a VSWR = 6.47 for 1850 MHz (the lowest frequency), VSWR min = 1.00 for 1922 MHz (the resonant frequency), VSWR = 1.00 for 1990 MHz (the highest frequency). The VSWR is at its minimum, it's a very interesting result. Also, The PCS bandwidth (140 MHz) is for the designed antenna a 1:6.47 VSWR bandwidth and the antenna presents a 1:2 bandwidth equal to 63 MHz.

2.6.2. The Impedance in the Feeding Point

The Figure 8 shows a regular impedance smith chart with interesting parameters of reflection, impedance, VSWR and Q. We can see very close values between the impedances (input and port). The feeding point position is then confirmed that is very adequate because it presents a very interesting adaptation.

2.6.3. The Antenna Parameters

The simulations results are summarized in Table 1. The obtained gain G is 1 dBi and the radiation efficiency is 1.0085.

2.6.4. The Diagram Pattern

We can confirm by the Figure 9 that (xz) is the E-plane. Also, by the tables the E-field has its maximum in ($\phi = 0$ deg and $\theta = -36$ deg).

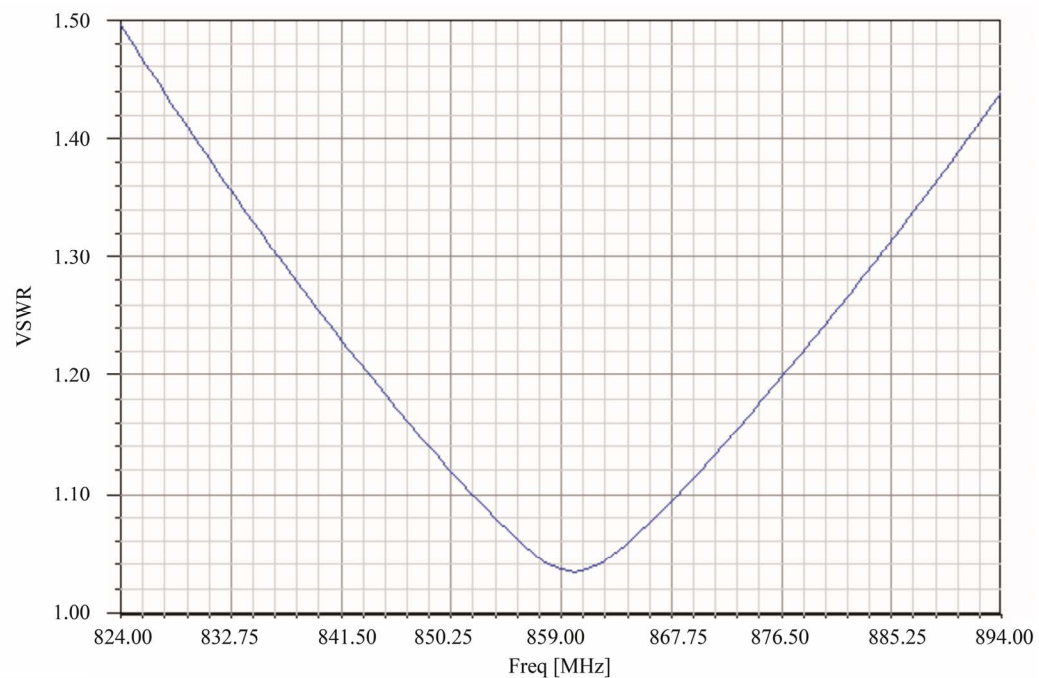
3. Comparison between the Three Scenarios

The mean results are shown by Table 2. We can detect that the second scenario gives more interesting results, the simultaneous parametric simulations increase the probability to find an optimized configuration.

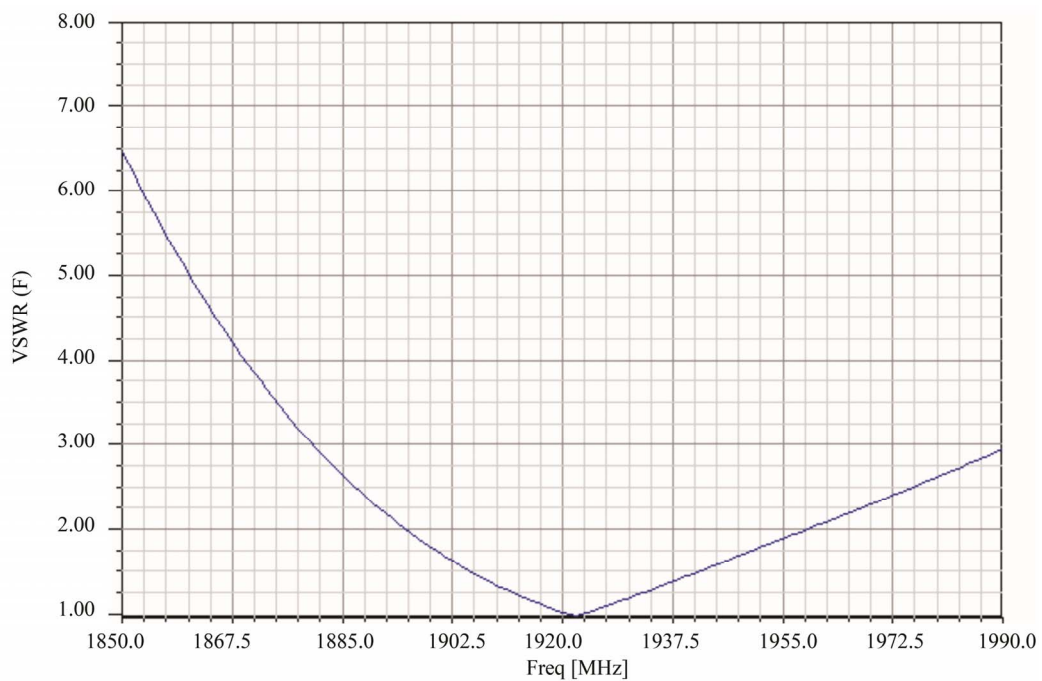
4. The Selected Methodology

By the precedent comparison, the author concluded about the following methodology that is improvement of the second scenario. To design antenna, we must group by the parameters that have the same order of magnitude to make simultaneous parametric simulation and also varying successively the parameters which have different order of magnitude. For this, the following methodology is efficient to design optimized multiband PIFAs:

- Choose the patch dimensions and shape by applying the theoretical formula of resonant frequency.
- Vary h from some millimeters to the maximum ac-



(a)



(b)

Figure 7. The VSWR depending on the frequency for both bands.

cepted thickness of the handset (in the hypothesis of an infinite ground plan that can be chosen in the simulation by HFSS). The height h is then selected. for multiband antenna, the optimal value should be a trade-off for different bands.

- Vary the ground plan dimensions regarding the maxi-

mum dimensions of the PCB card where the antenna will be mounted. The ground plan dimensions are then chosen.

- Make a simultaneous parametric simulations for the parameters concerning the feeding point and the shorting plate (as W_s , D , p). This step is compliant for

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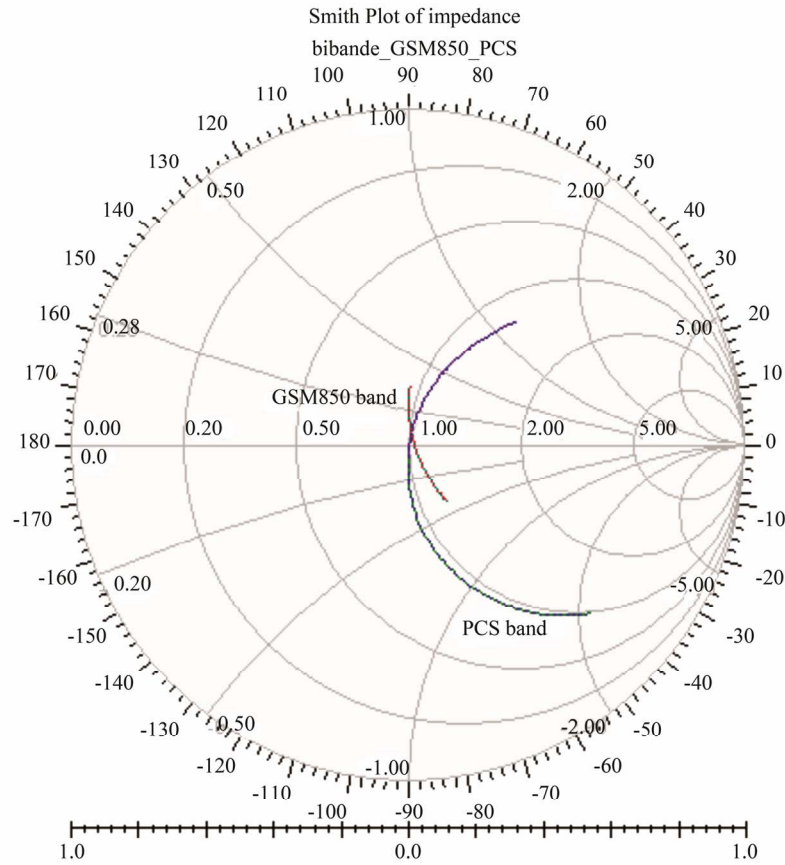


Figure 8. The impedance smith chart for both bands.

Table 1. The design antenna parameters and dimensions.

Antenna parameter	Value of the parameter GSM850 PCS1900
Patch Length L_p	31 mm
Patch width W_p	70 mm
Ground plan length L_g	90 mm
Ground plan width W_g	70 mm
Height h	11 mm
Short plate width W_s	3 mm
Short plate position D	30 mm
Feeding point position	3 mm
Resonant frequency	860 MHz, 1922 MHz
Peak S_{11}	-36 dB, -56 dB
1:2 VSWR bandwidth	63 MHz, 40 MHz
Peak Gain	1 dB
Peak directivity	1 dB
Radiation efficiency	85%
E total max	Phi = 0 deg, Theta = -36 deg

Table 2. Comparison of the three scenarios.

Antenna parameter	Scenario 1 Single Band	Scenario 2 First band Second band	Scenario3 First band Second band
Required bandwidth	140	70, 170	70, 140
-10 db bandwidth	> 140 MHz	>70 MHz, 65 MHz	>70 MHz, 40 MHz
1:1.5 VSWR bandwidth	130 MHz (93%)	>90 MHz (>130%), 40 MHz	70 (100%), 35 MHz
Peak S_{11}	-65 dB	-55 dB, -23 dB	-36 dB, -56 dB
Peak gain	1.16 dB	12.9 dB	1 dB
Peak directivity	1.15 dB	15.2 dB	1 dB
Radiation efficiency	1.0085	0.84	1 dB
Required resonant frequency	1920 MHz	925 MHz, 1795 MHz	859 MHz, 1920 MHz
Resonant frequency	1924 MHz	926 MHz, 1804 MHz	860 MHz, 1922 MHz

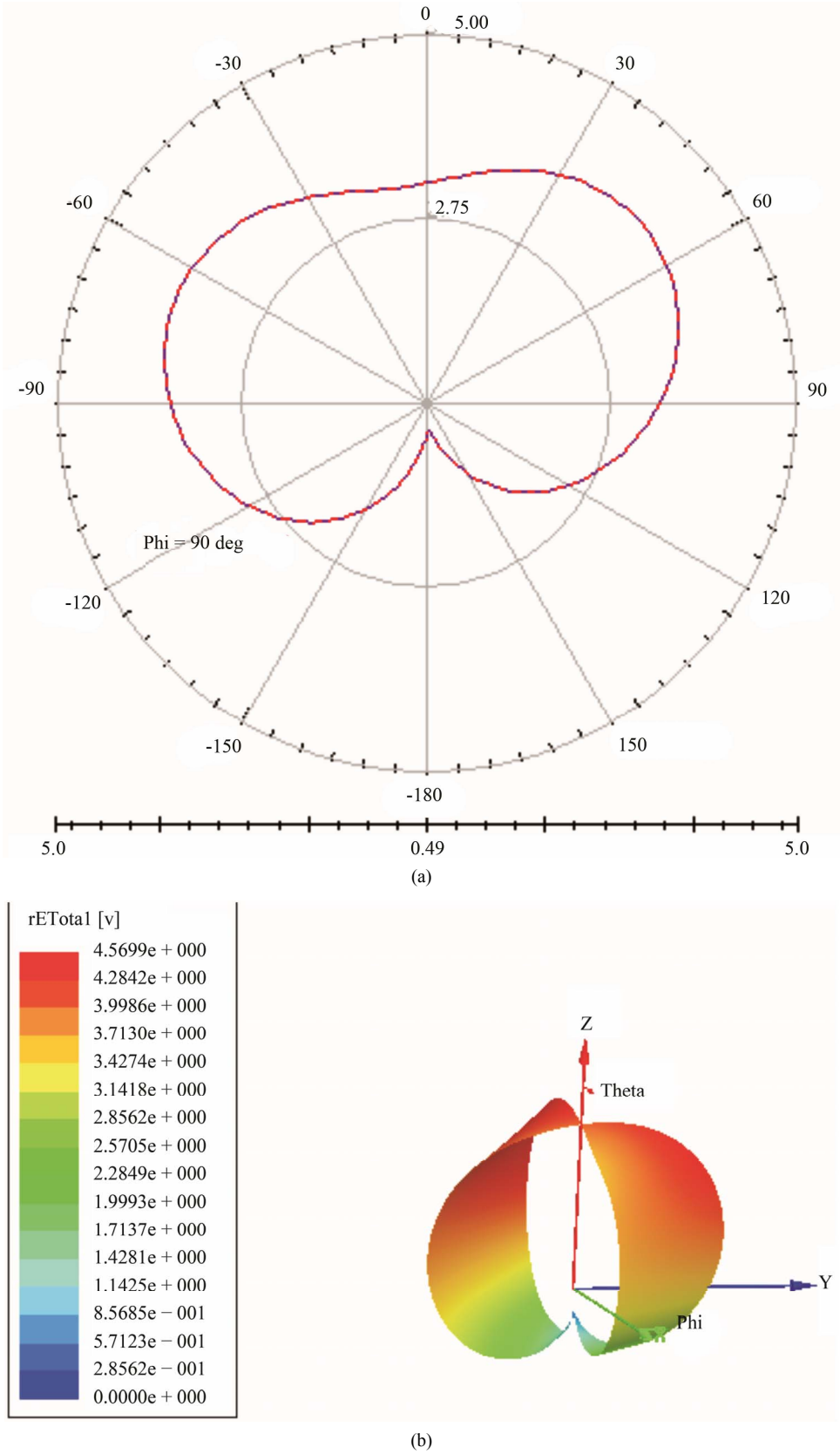


Figure 9. The E-field polar diagram pattern; (a) in 2D; (b) in 3D.

both antenna feeding ways (probe or plate). Make a trade off choice of the optimal configuration for different bands.

- This methodology is interesting not only to design PIFA with regular shape (as studied here), but also PIFAs with bent and meandered shapes by modifying an initial regular PIFA.

5. Conclusions

To design a dual band (or multibands) PIFA, the author exposed a methodology based on parametric simulations. The parameters variations are simulated successively or simultaneously by group that have the same order of magnitude. The methodology is a formalism of precedent published works of the author concerning the parametric simulations to design PIFAs. The results are interesting and the methodology can be used lonely or can be also combined with other design algorithms of PIFA literature.

Also, the methodology is a trade-off between different bands of the antenna because enhancement in a band affects negatively the other bands. The methodology has as goal to search an effective solution that respects the requirements and not necessary the optimal one. In comparison with precedent designed PIFAs [4,5] for different frequency bands, the methodology allows the design of antennas presenting a very high gain and directivity and also an interesting radiation efficiency for different band. The use of methodology in designing antenna is so interesting to optimize the antenna characteristics and performance. This research field is in progress and the antenna design methodology can be combined with algorithms as genetic algorithms to make more optimized antennas.

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